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ILLINOIS UNIV AT URBANA DEPT OF ELECTRICAL ENGINEERING
A FORTRAN PROGRAM FOR RECTANGULAR MICROSTRIP ANTENNAS. (U)

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APR 82 W F RICHARDS, Y T LO
UILU-ENG-80-2538

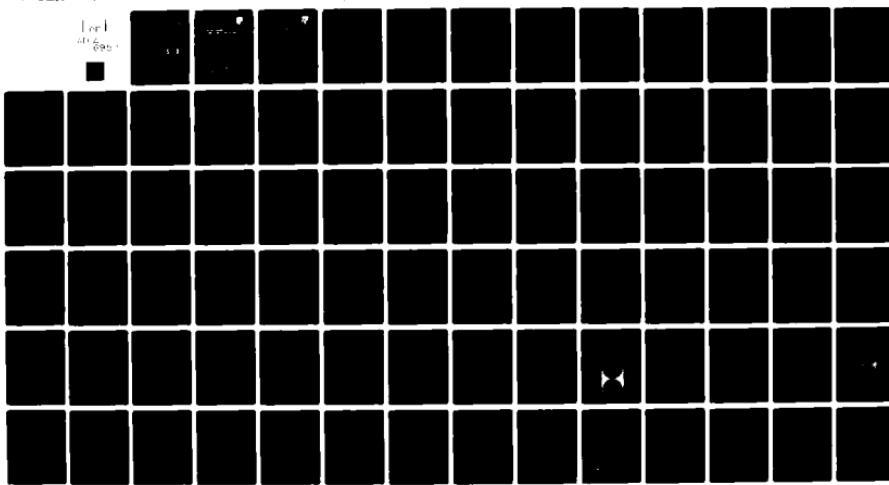
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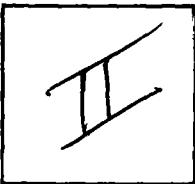


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Interim Report

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University of Illinois at Urbana-Champaign

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**ROME AIR DEVELOPMENT CENTER
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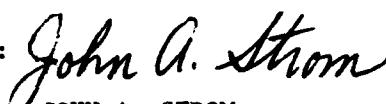
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RADC-TR-82-78 has been reviewed and approved for publication.

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INTRODUCTION

This report supplies a program, with examples, for the analysis of rectangular microstrip antennas. The formulas upon which the program is based are also provided. The theory from which these formulas were obtained is based on the "cavity model" of the microstrip antenna developed at the University of Illinois by Lo, Richards, *et al.* Details of the theory can be found in the references listed in the bibliography at the end of the report.

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CHAPTER 1: FORMULAS AND DEFINITIONS

For completeness, the basic formulas and procedures used to compute the pattern and impedance of a microstrip antenna are reviewed. The geometrical parameters used in the formulas that follow are defined in figure 1.

I GENERAL NOTATIONS

The following notations are used in the formulas within this appendix.

- (1) $k_0 = 2\pi f/c$ where f is the frequency and c is the speed of light in free space.
- (2) $\eta_0 = 377\Omega$.
- (3) $\epsilon_{0m} = 1$ for $m = 0$ and 2 otherwise.
- (4) (r, θ, ϕ) is the coordinate of a point in spherical coordinates. The direction perpendicular to the ground plane corresponds to $\theta = 0$. The line $\theta = \pi/2, \phi = 0$ is the x axis while $\phi = \pi/2$ is the y axis.
- (5) $p_m = (k^2 - (m\pi/a)^2)^{1/2}$ (the branch is irrelevant).
- (6) $k = k_0\epsilon^{1/2}(1-\delta)^{1/2}$ where the branch is also irrelevant, ϵ , is the relative dielectric constant of the dielectric substrate, and δ is the loss tangent of the dielectric substrate (later to be replaced by the "effective loss tangent." See V.)
- (7) $\Phi_m^{(1)} = \left(\frac{\epsilon_{0m}}{a}\right)^{1/2} \cos[p_m(b-y)]\cos(m\pi x/a)$, for $y \geq y_1$, and
- $\Phi_m^{(2)} = \left(\frac{\epsilon_{0m}}{a}\right)^{1/2} \cos(p_m y)\cos(m\pi x/a)$, for $y < y_1$.
- (8) $j_0(x) = \sin(x)/x$ (the spherical Bessel function of zero order).
- (9) Δ is the skin depth.

II PATTERN AND RADIATED POWER

Radiated power, P_{rad} , is computed in subroutine VRAD. This routine calls the double integration routine, VDOUBL, which applies 4-point Gaussian quadrature recursively to integrate the power pattern supplied by VPPAT. The function VPPAT calls VPAT which computes the complex polar pattern of the antenna by application of the following formulas:

$$F = -\frac{e^{-jk_0r}}{r} \frac{jk_0\eta_0 b}{2\pi} \sum_{m=0}^{\infty} \frac{\epsilon_{0m}\cos(m\pi x_1/a)}{p_m b \sin(p_m b)} J_0\left(\frac{m\pi d}{2a}\right) \left[(-1)^m e^{jk_0 a \sin\theta \cos\phi} - 1 \right] \\ \cdot \left\{ \hat{x} \left[\cos(p_m y_1) e^{jk_0 b \sin\theta \cos\phi} - \cos(p_m(b-y_1)) \right] \frac{j k_0 a \sin\theta \cos\phi}{(m\pi)^2 - (k_0 a \sin\theta \cos\phi)^2} \right\}$$

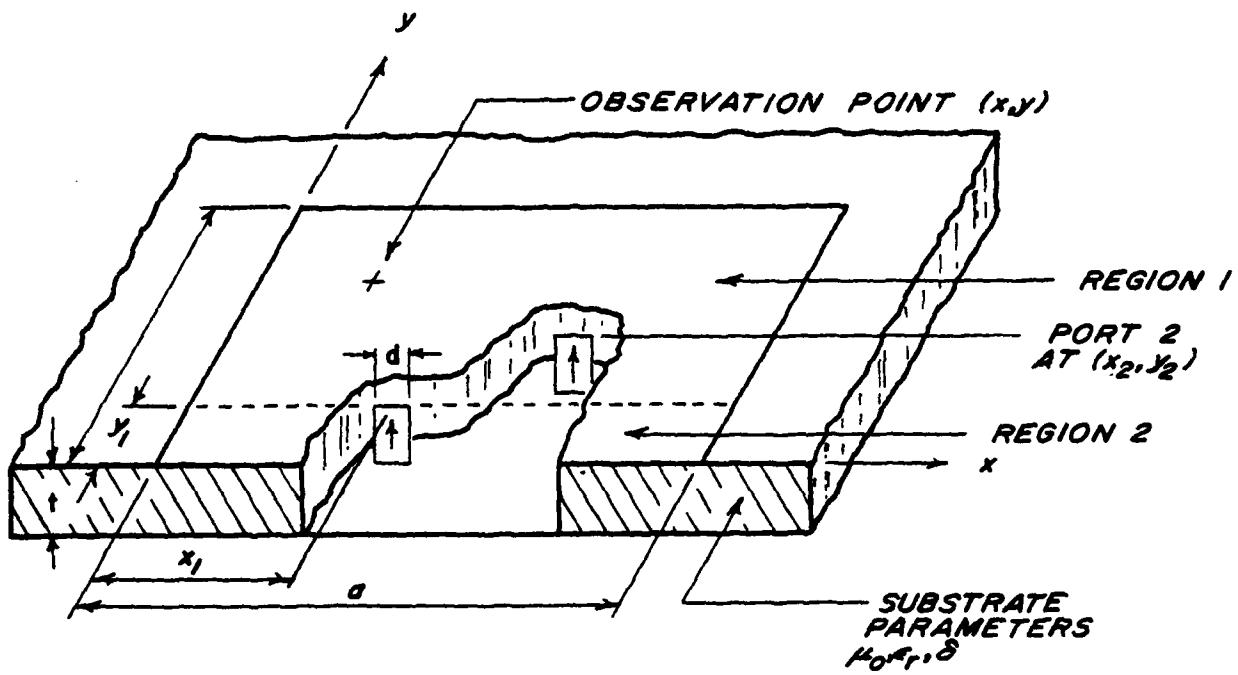


Figure 1. Geometry and Idealized Feeds for the Rectangular Microstrip Antenna

$$\begin{aligned}
& - \hat{y} \frac{b}{a} \left[p_m b \sin(p_m b) e^{jk_0 y_1 \sin \theta \sin \phi} + j k_0 b \sin \theta \sin \phi \right. \\
& \cdot \left. \left[\cos(p_m y_1) e^{jk_0 b \sin \theta \sin \phi} - \cos(p_m(b-y_1)) \right] \right] \\
& \cdot \left. \left[(p_m b)^2 - (k_0 b \sin \theta \sin \phi)^2 \right]^{-1} \right\}
\end{aligned}$$

$$E_\theta = k_0 (-F_x \sin \phi + F_y \cos \phi)$$

$$E_\phi = -k_0 (F_x \cos \phi + F_y \sin \phi) \cos \theta$$

where E_θ and E_ϕ are the θ and ϕ components of the electric far field, F_x and F_y are the x and y components of \vec{F} , and \hat{x} and \hat{y} are the unit vectors in the x and y directions.

III STORED ENERGY AND OHMIC LOSSES

The stored electric energy, W_E , is computed from

$$\begin{aligned}
4\pi f W_E &= \epsilon_r k_0 t (k_0 b)^2 \eta_0 \sum_{m=0}^{\infty} \frac{j_0^2 \left(\frac{m\pi d}{2a} \right)}{|p_m b \sin(p_m b)|^2} \\
&\cdot \left\{ y_1 \left| \Phi_m^{(1)}(x_1, y_1) \right|^2 N(p_m y_1) + (b-y_1) \left| \Phi_m^{(2)}(x_1, y_1) \right|^2 N[p_m(b-y_1)] \right\},
\end{aligned}$$

where

$$N(z) = \frac{1}{2} [j_0(j2\text{Im}z) + j_0(2\text{Re}z)].$$

The dielectric loss is found from

$$P = 4\pi f \delta W_E.$$

The copper loss is determined using

$$\frac{P_{Cu}}{P_d} \approx \frac{\Delta}{\delta t}$$

(at resonance). All these quantities are computed within subroutine VENLS.

IV IMPEDANCE

The impedances are computed in subroutines VZ1 and VZ2. The former is called by VZ2 to compute z_{11} and z_{22} while z_{12} is computed within VZ2 by the following formula for $y_2 > y_1$:

$$z_{12} = -jk_0 t \eta_0 \sum_{m=0}^{\infty} \left\{ \frac{\epsilon_{0m}}{a} \cos(m\pi x_1/a) \cos(m\pi x_2/a) j_0 \left(\frac{m\pi d}{2a} \right) \right\}$$

$$\frac{\cos[p_m(b-y_2)]\cos(p_my_1)}{p_m \sin(p_m b)} \Bigg\}.$$

For $y_1 = y_2$, this series is accelerated by writing it as

$$z_{12} = -jk_0 t \eta_0 \left\{ \frac{\cos(ky_1)\cos[k(b-y_1)]}{k \sin(kb)} \right. \\ + \sum_{m=1}^{\infty} \frac{2}{a} \cos(m\pi x_1/a) \cos(m\pi x_2/a) j_0^2 \left(\frac{m\pi d}{2a} \right) \\ \cdot \left[\frac{\cos(p_my_1)\cos[p_m(b-y_1)]}{p_m \sin(p_m b)} + \frac{a\tau}{m\pi} \right] \\ + \frac{jk_0 t \eta_0 \tau}{\pi^3} \left(\frac{a}{d} \right)^2 \left\{ F \left(\frac{\pi(x_1+x_2)}{a} \right) + F \left(\frac{\pi(x_1-x_2)}{a} \right) \right. \\ \left. - \frac{1}{2} \left[F \left(\frac{\pi(x_1+x_2+d)}{a} \right) + F \left(\frac{\pi(x_1+x_2-d)}{a} \right) + F \left(\frac{\pi(x_1-x_2+d)}{a} \right) + F \left(\frac{\pi(x_1-x_2-d)}{a} \right) \right] \right\} \right\}$$

where $\tau = 1$ for $b > y_1 > 0$ and $\tau = 2$ for $y_1 = 0$ or $y_1 = b$. The driving point impedance, z_{11} , is computed using the accelerated formula for z_{12} with x_2 and y_2 replaced by x_1 and y_1 , respectively. Similarly, z_{22} is computed with x_1 and y_1 replaced by x_2 and y_2 .

The function $F(x)$ is related to Clausen's integral and is given by

$$F(x) = \sum_{m=1}^{\infty} \cos \frac{(mx)}{m^3}.$$

This function is written in terms of $\ln x$ and a rapidly converging series of Chebyshev polynomials. It is evaluated in function VF.

V EFFECTIVE LOSS TANGENT

The "effective" loss tangent is found by first computing the fields within the "cavity" based on k found from the *actual* loss tangent of the substrate. From these fields, computations of the electric stored energy, the radiated power, and the power loss in the dielectric and copper are made. From these quantities, the antenna " Q " is computed from

$$Q = \frac{4\pi f W_E}{P_{\text{rad}} + P_d + P_{\text{Cu}}}.$$

An "effective" loss tangent, δ_{eff} , is defined as

$$\delta_{\text{eff}} = \frac{1}{Q}.$$

This loss tangent is then used to compute an improved k and the whole process is repeated to find new (and more accurate) predictions of the stored energy and losses. A new δ_{eff} is found, and so on. The program as supplied computes a twice iterated δ_{eff} . However, for thin substrates, the procedure converges after a single iteration and the first δ_{eff} computed is adequate. A simple modification of the program will eliminate the second iteration.

CHAPTER 2: PROGRAM LISTING

The FORTRAN program listed in this chapter was implemented on the CYBER 175 computer located at the University of Illinois, Urbana, IL. The program uses CDC's "extended FORTRAN" and the Graphics Compatibility System (GCS) produced by the United States Military Academy. Names of subroutines provided by GCS all begin with the letter "U." These GCS routines are used in certain input/output subroutines including those that plot results. Such routines have been listed below in a section entitled "INPUT/OUTPUT AND POTTING." It is this section of the program which is rather strongly installation dependent and would probably require user modification.

The other major sections of the program listed below are the "MAIN PROGRAM" and the "NUMERICAL" sections. The former controls the flow of execution of the program while the latter computes the impedance, pattern, etc. Both these sections are fairly transportable, particularly the NUMERICAL section. Only a few non-ANSI FORTRAN statements and routines are used in these sections and these can be easily eliminated or modified.

The overall simplified flow-chart of the program is shown in fig. 2. Other details of specific subroutines can be obtained by referring to their respective documentation in comment cards.

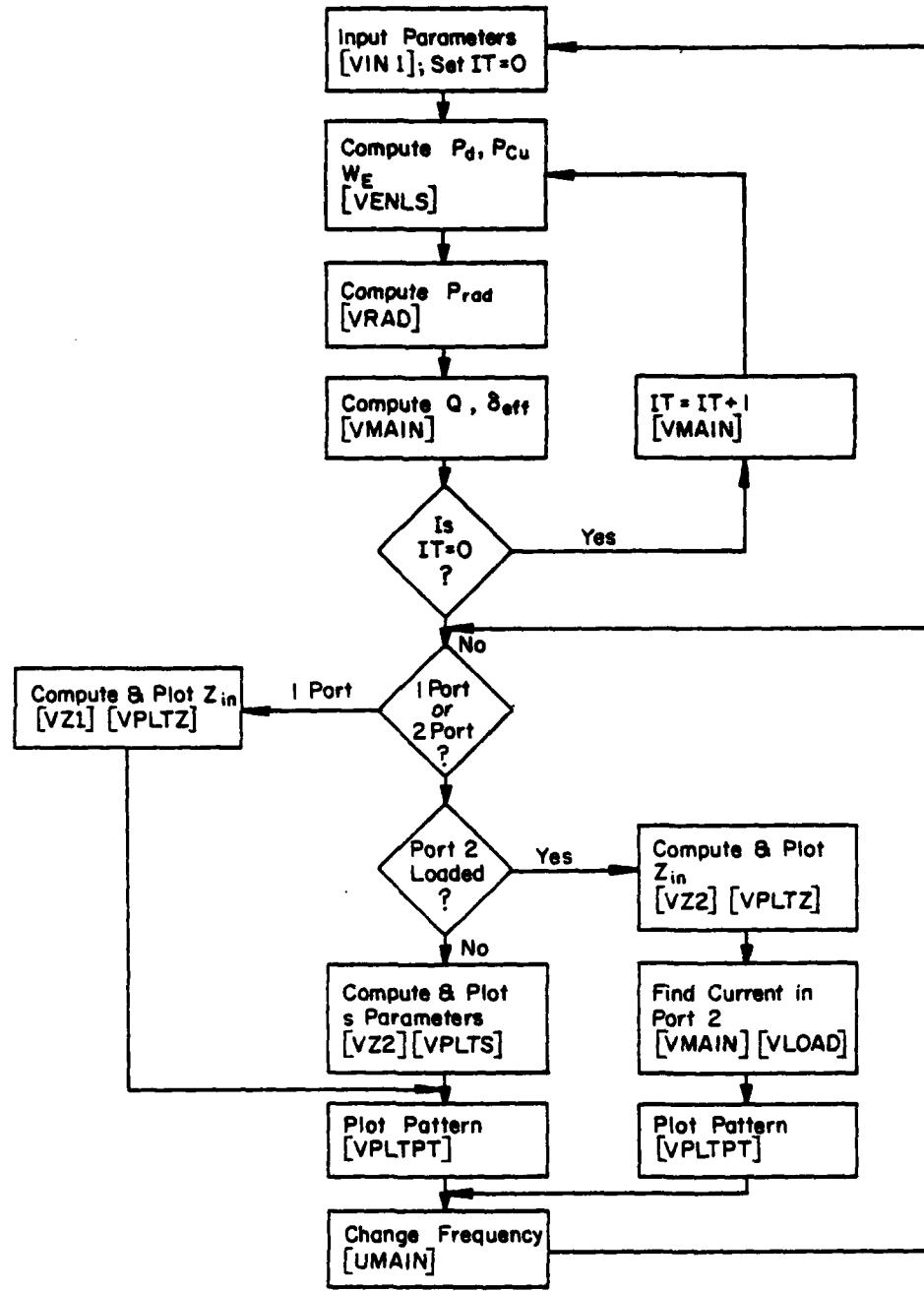


Figure 2. Simplified Flow Chart for the Program

MAIN PROGRAM

```
PROGRAM VMAIN(INPUT,OUTPUT,TAPE1=OUTPUT,TAPE2=INPUT,RESULT,  
1 TAPE3=RESULT) N/S  
*****  
** This is the main program of a group of routines that computes  
** (1) Input impedance of a  
**     (a) Single port rectangular microstrip antenna;  
**     (b) Two port rectangular microstrip antenna with one of its  
**         ports loaded by a specified impedance (in subroutine  
**         "VLOAD");  
** (2) The "s" parameters of a two port microstrip antenna;  
** (3) The radiation pattern of a  
**     (a) Single port rectangular microstrip antenna;  
**     (b) Two port rectangular microstrip antenna with one of its  
**         ports loaded by a specified impedance.  
*****  
* REFERENCES: The method used is described in the following  
* publications:  
*[1] Y. T. Lo, D. Solomon, W. F. Richards, "Theory and Experiment on  
* Microstrip Antennas," IEEE TRANS. ANTENNAS PROPAGAT. Vol.  
* AP-27, pp. 137-145, MAR 79.  
*[2] Y. T. Lo, W. F. Richards, D. D. Harrison, "An Improved Theory  
* for Microstrip Antennas and Applications," RADC-TR INTERIM  
* REPORT (PART I), DEC 78.  
*[3] W. F. Richards, Y. T. Lo, D. D. Harrison, "Improved Theory  
* for Microstrip Antennas," IEE ELECTRONICS LETTERS, Vol. 15,  
* pp. 42-44, JAN 79.  
* LIMITATIONS:  
* The current version does not include an estimate of surface wave  
* power as this computation is currently under critical evaluation.  
* This version also requires the specification of the so called  
* "effective feed width." This parameter arises from an attempt to  
* idealize the fields in the vicinity of a coaxial or microstrip feed  
* so that the source can be considered as a uniform current  
* ribbon of width D (the effective width) flowing from the patch to  
* the ground plane. Since the observed shift of impedance loci into  
* inductive half of the Smith Chart depends rather strongly on the  
* field distribution in the vicinity of the feed, this idealization  
* needs some refinements and a more rigorous treatment of this  
* problem is under way. For the present, the user should try some  
* different values of D until he finds one which fits measured results  
* most closely. The representation of the fields for frequencies far  
* away from resonance, say near the mean of two widely spaced adjacent  
* resonant frequencies, is currently not sufficiently accurate for all  
* applications. We will do further research to develop better compu-  
* tations in this regime.  
* IMPLEMENTATION REQUIREMENTS: Except for the input/output which  
* relies heavily upon the graphics capabilities of the GCS system  
* developed by the United States Military Academy, the program is  
* written in ANSI FORTRAN and should be relatively transportable.  
* All GCS subroutine names begin with a "U" in this program.  
* Some of the input/output utilizes extended features of CDC's FORTRAN  
* as implemented on the University of Illinois' CYBER 175 (NOS V. 4.7)  
* and will have to be modified for use on other systems.  
* Non-ANSI FORTRAN statements are flagged as N/S.  
* USER INSTRUCTIONS: The parameter descriptions and options are  
* explained through the use of examples provided with this listing.  
*****
```

```

REAL LOSS, K0, LOSSO
INTEGER P, ANS, PO
COMPLEX Z, Z11, Z12, Z22, ZL, ZIN, I2
COMMON /DELTA/ DELTA
COMMON /OPT/ ANS
COMMON /I1/ A, B, T, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,
          F0, NFREQ, K0, TWOPI, PDO2A, PXPOA, BOA, BAND, PI,
          ETAO, K, GAIN, DELTAF
COMMON /I2/ AO, B0, T0, DO, DIELO, LOSSO, SIGMO, PO, XPO, YPO, XPP0, YPP0,
          LO, FOO, BAND0, DELTA0
DATA AO/0./, B0/0./, T0/0./, DO/0./, DIELO/1./, LOSSO/0./, SIGMO/580./,
      PO/1., XPO/0./, YPO/0./, XPP0/0./, YPP0/0./, LO/"N"/, FOO/0./,
      BAND0/0./, DELTA0/0.0/, C/30000./, PI/3.1415926535898/,
      TWOPI/6.2831853071796/, ETAO/377./
CALL USTART
1 CALL VIN1
IT = 0
DELTA = LOSS
C
C Find the nearest resonant frequency to the specified center frequency
C
CALL VSEARC (M, N)
K0 = SQRT(((M*PI/A)**2 + (N*PI/B)**2)/DIEL)
PDO2A = PI * D / (2 * A)
PXPOA = PI * XP / A
BOA = B / A
C
C Compute "closed form" sum of asymptotic expression of the summand
C for the driving point impedance series
C
S1 = VS(XP, XP)
IF (P .LT. 2) GO TO 2
S2 = VS(XPP, XPP)
S3 = VS(XP, XPP)
C
C Compute stored electric energy, WWE, copper loss, PCU, and dielectric
C loss, PD, for "effective loss tangent", DELTA. (Dielectric loss is
C always computed using the actual loss tangent, LOSS).
C
2 CALL VENLS (DELTA, WWE, PCU, PD)
C
C Compute radiated power, PRAD, for effective loss tangent, DELTA
C
CALL VRAD (DELTA, PRAD)
POWER = PRAD + PCU + PD
Q = 2 * WWE / POWER
DELTA = 1/Q
IT = IT + 1
C
C Iterate the calculation twice to ensure proper value of DELTA
C is obtained.
C
IF (IT .LT. 2) GO TO 2
C
C Compute the pattern along the zenith direction to determine
C antenna gain.
C
CALL VPAT (0., 0., Z11, Z22, DELTA, (1., 0.), XP, YP)
E = Z11*CONJG(Z11)+Z22*CONJG(Z22)
GAIN = 10.* ALOG(2*TWOPI*E/(ETAO*PRAD))
1   /ALOG(10.)
F = F0 - (NFREQ/2)*DELTAF
NO2 = NFREQ/2 + 1
WRITE (1,7)
READ (2,8) ANS
CALL EOF(2)
IF (ANS .EQ. 0) ANS = 3
DO 3 K = 1, NFREQ
  K0 = TWOPI * F / C

```

N/S

```

C          GO TO (10, 20), P
CC         Find the driving point impedance of the one port
C
C 10      CALL VZ1 (Z, DELTA, XP, YP, S1)
CC         Input data to the plotting program, VPLTZ
C
C 20      CALL VPLTZ (Z, F, 0)
CC         GO TO 30
C
C         Compute the "z" parameters of the two port
C
C 20      CALL VZ2 (Z11, Z12, Z22, DELTA, S1, S2, S3)
IF (L .EQ. "N") GO TO 202
C
C         Compute impedance of load on port two of the microstrip
C
C 20      CALL VLOAD (F, ZL)
C
C         Compute the input impedance as seen at port one of the loaded
CC         microstrip antenna.
C
C 20      ZIN = Z11 - Z12**2 / (Z22 + ZL)
CALL VPLTZ (ZIN, F, 0)
C
C         Compute the current flowing through the load at port two.
C
C 20      I2 = -Z12 / (Z22 + ZL)
C
C         Depending on the options chosen, compute the pattern of the antenna.
C
C 20      IF (ANS .EQ. 1 .OR. (ANS .EQ. 2 .AND. K .EQ. NO2))
1          CALL VPLTPT (I2, F)
GO TO 3
C
C         Input the two port parameters to the "s" parameter plotting
CC         program for the case of a non-loaded two port antenna.
C
C 20      CALL VPLTS (F, Z11, Z12, Z22, 0)
30     IF (ANS .EQ. 1 .OR. (ANS .EQ. 2 .AND. K .EQ. NO2))
1          CALL VPLTPT (I2, F)
3          F = F + DELTAF
IF (P .EQ. 1) GO TO 5
IF (L .NE. "N") GO TO 5
C
C         Plot "s" parameters
C
C 20      CALL VPLTS (F0, Z11, Z12, Z22, 1)
GO TO 6
C
C         Plot the input impedance to the microstrip.
C
C 5      CALL VPLTZ (Z, F0, 1)
6      WRITE (1,9)
READ (2,1) ANS
CALL EOF(2)
IF (ANS .EQ. 0) ANS = "Y"
IF (ANS .EQ. "Y") GO TO 1
STOP
7      FORMAT ("Choose an option: ",/
, "(1) Plot patterns at",
2" all frequencies.",/
, "(2) Plot pattern only",
3" at center frequency.",/
, "(3) Plot no patterns",/
, "Type option (1, 2, or 3): ",)
8      FORMAT(11)
9      FORMAT ("Continue (type Y or N) ",)

```

N/S

11 FORMAT (A1)
END

NUMERICAL

```

SUBROUTINE VDOUBL (F, M, A, B, C, D, INT)
*****  

* PURPOSE: This subroutine performs the integral from 0 to A of the  

* integral from 0 to B of the externally declared function  

* F(X,Y).  

* PARAMETERS: The parameters are as stated above and M is the log base  

* two plus one of the number of point Gaussian quadrature  

* formula used in the mechanical quadrature. INT is the  

* integral.  

*****  

REAL INT, W(31), S(31)
DATA W / 1.000000000000000, .65214515486255, .34785484513745,  

* .36268378337836, .31370664587789, .22238103445337,  

* .10122853629038, .18945061045507, .18260341504492,  

* .16915651939500, .14959598881658, .12462897125553,  

* .09515851168249, .06225352393865, .02715245941175,  

* .09654008851473, .09563872007927, .09384439908080,  

* .09117387869576, .08765209300440, .08331192422695,  

* .07819389578707, .07234579410885, .0658222277636,  

* .05868409347854, .05099805926238, .04283589802223,  

* .03427386291302, .02539206530926, .01627439473091,  

* .00701861000947/  

DATA S / .57735026918963, .33998104358486,  

* .86113631159405, .18343464249565, .52553240991633,  

* .79666647741363, .96028985649754, .09501250983764,  

* .28160355077926, .45801677765723, .61787624440264,  

* .75540440835500, .86563120238783, .94457502307323,  

* .98940093499165, .04830766568774, .14447196158280,  

* .23928736225214, .33186860228213, .42135127613064,  

* .50689990893223, .58771575724076, .66304426693022,  

* .73218211874029, .79448379596794, .84936761373257,  

* .89632115576605, .93490607593774, .96476225558751,  

* .98561151154527, .99726386184948/  

L1 = 2**(M-1)
N1 = 2*L1 - 1
F1 = (B-A)/2
F2 = (D-C)/2
F3 = (B+A)/2
F4 = (D+C)/2
INT = 0.
DO 2 J = L1, N1
    SUM = 0.
    DO 1 K = L1, N1
        T1 = F1 * S(K)
        T2 = F2 * S(J)
        T3 = T1 + F3
        T4 = T2 + F4
        T5 = F3 - T1
        T6 = F4 - T2
1    SUM = SUM+W(K)*(F(T3,T4)+F(T3,T6)+F(T5,T4)+F(T5,T6))
2    INT = INT + W(J) * SUM
    INT = INT * F1 * F2
RETURN
END

```

```

SUBROUTINE VENLS(DELTA,ESE,DL,CL)
*****  

* PURPOSE: This subroutine computes the electric stored energy and the  

* copper and dielectric losses at resonance.  

*  

* PARAMETERS: DELTA is the effective loss tangent. ESE is the computed  

* electric stored energy (at the nearest resonance to F0 as  

* determined by subroutine VSEARC). DL is the dielectric loss  

* computed using the actual loss tangent. CL is the copper loss  

* computed as a proportion of the dielectric loss.  

*****  

REAL MPI,MPIOA,MPIOA2,LT,JO,N,K0
COMPLEX K,SPB,P,PYP,PSPB,PBMYP,CPYP,CPBMYP
COMMON /I1/A,B,T,D,EPSS,LTSIGM,IP,XP,YP,XPP,YPP,L,
F0,NFREQ,K0,TWOP1,PDO2A,TPXPOA,BOA,BAND,P1,
ETA,KK,GAIN,DELTAF
K = CSQRT(CMPLX(EPS,-DELTA*EPS)) * K0
SD = SQRT(2/(1000*K0*ETA*SIGM))
BMYP=B-YP
CPYP=CCOS(K*YP)
CPBMYP=CCOS(K*BMYP)
SPB=CSIN(K*B)
F1=YP*(CABS(CPBMYP))**2
F2=BMYP*(CABS(CPYP))**2
ESE=(F1*N(K*YP)+F2*N(K*BMYP))/(CABS(K*SPB))**2
M1=0
100 CONTINUE
M1=M1+1
MPI=M1*3.141592654
MPIOA=MPI/A
MPIOA2=MPIOA**2
P=CSQRT(K*K-MPIOA2)
PSPB=P*CSIN(P*B)
PYP=P*YP
PBMYP=P*BMYP
CPYP=CCOS(PYP)
CPBMYP=CCOS(PBMYP)
CF=(JO(M1*PDO2A)*COS(MPI*XP/A))**2
F=2./(CABS(PSPB))**2
F1=YP*(CABS(CPBMYP))**2
F2=BMYP*(CABS(CPYP))**2
T3=(F1*N(PYP)+F2*N(PBMYP))*F
IF (MOD(M1,2) .EQ. 0) GO TO 150
SUBTOT = CF*T3
GO TO 100
150 SUBTOT = SUBTOT + CF*T3
ESE=ESE+SUBTOT
IF (SUBTOT/ESE .LT. 0.0001) GO TO 200
GO TO 100
200 ESE=ESE*EPS*T*K0**3*ETA/(2*A)
DL=2*LT*ESE
CL=SD*DL/(T*LT)
RETURN
END

```

```

FUNCTION VF(Z)
*****
* PURPOSE: This routine computes a sum related
* to the integral of Clausen's integral:
*
* Sum from k = 1 to infinity of cos(kZ)/k**3.
*
* METHOD: A Tchebyshev series was derived from
* expansions given in Abramowitz & Stegun
* and is summed by Clenshaw's algorithm.
*****
REAL C1(9), C2(9), LN202
DATA C1/.06905464409766968,
1   .00052071642034163,
2   .00000505847024742,
3   .0000006775156996,
4   .0000000107462004,
5   .0000000001896987,
6   .0000000000036087,
7   .000000000000726,
8   .000000000000015/
DATA C2/.2085520596900512,
1   .0029753216441284,
2   .0000406999817326,
3   .0000006926756791,
4   .0000000132725892,
5   .000000002743922,
6   .000000000059818,
7   .000000000001356,
8   .000000000000032/
TWOPI = 6.283185307179586
PIBY3 = 1.047197551196598
ZETA3 = 1.202056903159594
LN202 = 3465735902799726
X = ABS(Z)
IX = X / TWOPI
IF (X .LT. 0.) IX = IX - 1
Y = X - TWOPI * IX
IF (Y .GT. 3.141592653589793) Y = TWOPI - Y
KODE = 1
IF (Y .GE. 2.094395102393196) KODE = 2
IF (KODE .EQ. 2) Y = 3.141592653589793 - Y
T = Y / PIBY3
IF (KODE .EQ. 1) T = T / 2.
T2TSM1 = 2. * (2. * T**2 - 1.)
G1 = 0.
G2 = 0.
DO 4 L = 1, 9
  FACTOR = T2TSM1
  IF (L .EQ. 9) FACTOR = .5 * FACTOR
  G0 = FACTOR * G1 - G2
  GO TO {1,2} KODE
1  C = C1(10-L)
  GO TO 3
2  C = C2(10-L)
  G0 = G0 + C
  G2 = G1
  G1 = G0
4  CONTINUE
  G0 = G0 * (Y / TWOPI)**2
  GO TO {5,6}, KODE
5  VF = ZETA3
  IF (Y .NE. 0.) VF = VF + Y**2 * (0.5 * ALOG(Y) - 0.75 - 2.*G0)
  GO TO 7
6  VF = -0.75 * ZETA3 + Y**2 * (LN202 - 2. * G0)
7  RETURN
END

```

```

SUBROUTINE VLOAD(F, ZL)

* PURPOSE: This is an example of the format of user supplied subroutine
* VLOAD. The purpose of the subroutine is to return a load
* impedance, stored in ZL, at a frequency, F. Many of the
* electrical and geometrical parameters of the antenna are
* available to this routine through common block /I1/.
* Also, the parameter "K" is available through this common
* block. This parameter is a "DO" index for the loop that
* increments frequency in the main program. Thus, as was done
* in this example, and "IF" statement testing to see if K is
* one or not can be included so that data can be input by this
* program. Another common block, /LDID/ contains a string
* which allows one to give a verbal description of the type of
* load defined in the subroutine. The description must be 40
* characters or less in the CYBER system.

COMPLEX ZL
INTEGER STRING(4)
COMMON /I1/ A, B, T, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,
*          F0, NFREQ, K0, TWOPI, PDO2A, PXPOA, BOA, BAND, PI,
*          ETAO, K, GAIN, DELTAF
COMMON /LDID/ STRING
IF (K .GT. 1) GO TO 1
PRINT *, "INPUT CAPACITANCE IN PICOFARADS: C = ", N/S
READ *, C
ENCODE (40, 2, STRING) C
C = C * 1.E-6 N/S
1 ZL = CMPLX(0., -1./(F*TWOPI*C))
2 FORMAT ("CAPACITIVE LOAD: C = ", E8.2, ")"
RETURN
END

```

CCCCCCCC
* PURPOSE: This routine returns a load impedance of zero, (the
* impedance of a short). It is loaded as the defualt VLOAD
* subroutine.

SUBROUTINE VLOAD(F, ZL)
INTEGER STRING(4)
COMPLEX ZL
COMMON /LDID/ STRING
DATA STRING /"short circuit" /
ZL = (0.,0.)
RETURN
END

C *****
C *
C * PURPOSE: This group of functions compute certain quantities used in the
C * evaluation of electric stored energy computed in subroutine
C * VENLS.
C *****

REAL FUNCTION M (Z)
COMPLEX Z
REAL JO, IO
M = (IO(2*AIMAG(Z)) - JO(2*REAL(Z)))/2
RETURN
END
REAL FUNCTION N (Z)
COMPLEX Z
REAL JO, IO
N = (JO(2*REAL(Z)) + IO(2*AIMAG(Z)))/2
RETURN
END
REAL FUNCTION IO (X)
T = EXP(X)
IO = 1.
IF (X .NE. 0.) IO = (T - 1/T) / (2*X)
RETURN
END

```

SUBROUTINE VPAT (THETA, PHI, ETHETA, EPHI, DELTA, I, X, Y)
*****  

* PURPOSE: To evaluate the electric far field at direction  

* (THETA,PHI)  

*  

* PARAMETERS: ETHETA and EPHI are the THETA and PHI  

* components of electric field. DELTA is the  

* loss tangent used in the computation.  

*****  

* COMPLEX ETHETA, EPHI, FX, FY, PM, KSQ, PMB, PMY, CPY, CPBMY, I, XF,  

* F, PMBSPB, EKYSS, EKBSS, XT, YT, XTERM, YTERM, D2, FACTOR  

* REAL KO, MPI, KASC, KBSS, KYSS, JO  

* COMMON /I1/ A, B, T, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,  

* F0, NFREQ, KO, TWOPI, PDO2A, TPXPOA, BOA, BAND, PI,  

* ETAO, K, GAIN, DELTAF  

*  

* PXOA = PI * X / A  

FX = {0.,0.}  

FY = {0.,0.}  

M = 0  

CT = COS(THETA)  

ST = SIN(THETA)  

CP = COS(PHI)  

SP = SIN(PHI)  

KYSS = KO * Y * SP * ST  

KBSS = KO * B * SP * ST  

KASC = KO * A * CP * ST  

KSQ = DIEL * KO**2 * CMPLX(1., -DELTA)  

1 MPI = M * PI  

PM = CSQRT(KSQ - (MPI/A)**2)  

PMY = PM * Y  

PMB = PM * B  

CPY = CCOS(PMY)  

CPBMY = CCOS(PMB-PMY)  

PMBSPB = CSIN(PMB) * PMB  

EKYSS = CEXP(CMPLX(0., KYSS))  

EKBSS = CEXP(CMPLX(0., KBSS))  

D1 = MPI**2 - KASC**2  

D2 = PMB**2 - KBSS**2  

XF = CPY * EKBSS - CPBMY  

F = CEXP(CMPLX(0., KASC))  

IF (M .NE. 2*(M/2)) F = -F  

F = F - 1.  

IF (ABS(D1) .GT. 1.E-5) GO TO 2  

C >>>----> Find limiting value of expression when D1 ----> 0.  

C  

XT = XF  

IF (M .GT. 0) XT = XT / 2.  

GO TO 3  

2 XT = XF * F * CMPLX(0., KASC) / D1  

3 YT = -BOA * F * (PMBSPB * EKYSS + CMPLX(0., KBSS)*XF) / D2  

FACTOR = COS(M * PXOA) * JO(M * PDO2A) / (PMBSPB)  

IF (M .GT. 0) FACTOR = FACTOR * 2.  

M = M + 1  

IF (MOD(M,2) .EQ. 0) GO TO 30  

XTERM = FACTOR * XT  

YTERM = FACTOR * YT  

GO TO 1  

30 XTERM = XTERM + FACTOR * XT  

YTERM = YTERM + FACTOR * YT  

FX = FX + XTERM  

FY = FY + YTERM  

IF (SQRT(CABS(XTERM)**2 + CABS(YTERM)**2) .LT. 0.0001 *  

* SQRT(CABS(FX)**2 + CABS(FY)**2)) GO TO 4  

GO TO 1

```

```
4 FACTOR = K0**2 * T * ETA0 * B / TWOPI
ETHETA = (-FX*SP + FY*CP) * FACTOR * I
EPhi = -(FX*CP + FY*SP) * CT * FACTOR * I
RETURN
END

REAL FUNCTION JO(X)
JO = 1.
IF (X .NE. 0.) JO = SIN(X) / X
RETURN
END
```

```

FUNCTION VPPAT (THETA, PHI)

***** PURPOSE: THIS SUBROUTINE COMPUTES THE POWER PATTERN (TIMES THE SIN OF
***** THE POLAR ELEVATION ANGLE, THETA) FROM THE COMPLEX PATTERN
***** COMPUTED IN SUBROUTINE VPAT.

***** PARAMETERS: (THETA, PHI) IS THE DIRECTION OF OBSERVATION IN SPHERICAL
***** COORDINATES.

***** COMPLEX ETHETA, EPHI
***** COMMON /DELTA/ DELTA
***** COMMON /I1/ A, B, T, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,
***** F0, NFREQ, KO, TWOPI, PDO2A, TPXPOA, BOA, BAND, PI,
***** ETAO, K, GAIN, DELTAF
***** CALL VPAT (THETA, PHI, ETHETA, EPHI, DELTA, (1.0.), XP, YP)
***** VPPAT = SIN(THETA) * (ETHETA * CONJG(ETHETA) + EPHI * CONJG(EPHI))
***** RETURN
***** END

```

C
C
C
C
C
C
C
SUBROUTINE VRAD (DELTA, PRAD)

*
* PURPOSE: To evaluate the power radiated by the microstrip antenna
*
* PARAMETERS: DELTA is the effective loss tangent, and PRAD is the
* radiated power computed using numerical quadrature.
*

* COMMON /I1/ A, B, T, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,
* F0, NFREQ, KO, TWOPI, PDO2A, TPXPOA, BOA, BAND, PI,
* ETAO, K, GAIN, DELTAF
EXTERNAL VPPAT
CALL VDOUBL (VPPAT, 3, 0., PI/2., 0., TWOPI, PRAD)
PRAD = PRAD / ETAO
RETURN
END

```

FUNCTION VS (X1, X2)
=====
* PURPOSE: THIS FUNCTION EVALUATES THE CONTRIBUTION DUE TO THE FIRST
* TERM IN THE ASYMPTOTIC SERIES OF THE SUMMAND IN THE Z-PARAM
* EXPRESSIONS. (THIS IS USED TO APPLY KUMMER'S TRANSFORMATION
* TO ACCELERATE THE CONVERGENCE OF THE SERIES).
*
* PARAMETERS: X1 AND X2 ARE THE ABSCESSA OF THE LOCATIONS OF PORTS 1
* AND 2, RESPECTIVELY.
*
=====
COMMON /I1/ A, B, T, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,
* F0, NFREQ, KO, TWOPI, PDO2A, TPXPOA, BOA, BAND, PI,
* ETAO, K, GAIN, DELTAE
F1 = PI * (X1 + X2) / A
F2 = PI * (X1 - X2) / A
F3 = PI * (X1 + X2 + D) / A
F4 = PI * (X1 + X2 - D) / A
F5 = PI * (X1 - X2 + D) / A
F6 = PI * (X1 - X2 - D) / A
VS = VF(F3) + VF(F4) + VF(F5) + VF(F6)
VS = VF(F1) + VF(F2) - 0.5 * VS
VS = -(A/D)**2 * (1/PI)**3 * VS
RETURN
END

```

```

C SUBROUTINE VSEARC (MO, NO)
C ****
C * PURPOSE: This subroutine searches for the combination of mode indices,
C * (MO,NO) which yields the resonant wave number closest two the
C * wave number of free space at the chosen center frequency
C * times the permitivity of the dielectric.
C ****
C
REAL MIN, KG, KMN
INTEGER V
COMMON /I1/ A, B, TT, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,
* F0, NFREQ, KO, TWOP1, PDO2A, TPXPOA, BOA, BAND, PI,
* ETAO, K, GAIN, DELTAF
DATA C/30000./
KG = TWOP1 * F0 * SQRT(DIEL) / C
M = KG * A / PI
N = 0
MIN = (M+1) * PI / A - KG
MO = M + 1
NO = 0
V = 1
1 IF (V .EQ. 1) GO TO 2
M = M - 1
GO TO 3
2 N = N + 1
3 KMN = SQRT((M*PI/A)**2 + (N*PI/B)**2)
IF (KMN .LT. KG) GO TO 4
V = 0
GO TO 5
4 V = 1
5 T = ABS(KMN - KG)
IF (T .GE. MIN) GO TO 6
MIN = T
MO = M
NO = N
6 IF (M .NE. 0) GO TO 1
IF (V .EQ. 1) GO TO 7
N = N - 1
GO TO 8
7 N = N + 1
8 T = ABS(N*PI/B - KG)
IF (T .GE. MIN) RETURN
MO = 0
NO = N
RETURN
END

```

```

SUBROUTINE VZ1 (Z, DELTA, X, Y, S)
=====
#
# PURPOSE: This subroutine computes the driving point impedance of a
# rectangular microstrip antenna feed at point (X,Y).
#
# PARAMETERS: Z is the complex driving point impedance. DELTA is the
# effective loss tangent. (X,Y) is the coordinate of the feed.
# S is the "closed form" sum of the asymptotic form of the
# summand for Z. (It is used to accelerate the convergence
# the series.
#
COMPLEX Z, TERM, K, KSQ, PM, PMB, PY, PBMY, SUBTOT
REAL MPI, KO, LOSS, JO
COMMON /1/ A, B, T, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,
        FO, NFRÉQ, KO, TWOPI, PDO2A, PXPOA, BOA, BAND, PI,
        ETAO, KK, GAIN, DELTAF
TAU = 1.0
IF (Y .EQ. 0.) TAU = 2.
KSQ = DIEL * KO**2 * CMPLX(1.,-DELTA)
K = CSQRT(KSQ)
Z = A * S + CCOS(K*Y) * CCOS(K*(B-Y)) / (K*CSIN(K*B))
M = 0
SUBTOT = (0., 0.)
PXOA = PI * X / A
1 M = M + 1
MPI = M * PI
PM = CSQRT(KSQ - (MPI/A)**2)
PMB = PM * B
PY = PM * Y
PBMY = PMB - PY
TERM = (2*CCOS(PY)*CCOS(PBMY)/(PM*CSIN(PMB)) + TAU*A/MPI) *
       (COS(M*PXOA) * JO(M*PDO2A))**2
SUBTOT = SUBTOT + TERM
IF (3*(M/3) .NE. M) GO TO 1
Z = Z + SUBTOT
IF (CABS(SUBTOT) .LT. 0.001 * CABS(Z)) GO TO 2
SUBTOT = (0.,0.)
GO TO 1
2 Z = -Z * CMPLX(0.,1.) * KO * T * ETAO / A
RETURN
END

```

SUBROUTINE VZ2 (Z11, Z12, Z22, DELTA, S1, S2, S3)

```

* PURPOSE: This subroutine computes the open circuit parameters of the
* two port with port one at (XP,YP) and port two at (XPP,YPP)
* (where these parameters are in common block I1).
*
* PARAMETERS: Z11, Z12, and Z22 are the computed open circuit parameters
* DELTA is the effective loss tangent. S1, S2, and S3 are the
* "closed form" sums of the asymptotic form of the summands
* corresponding to Z11, Z12, and Z22 summations, respectively.
*
* COMPLEX Z11, Z12, Z22, TERM, K, KSQ, PM, PMB, PYP, PBMYPP, SUBTOT
* REAL MPI, K0, LOSS, J0
* COMMON /I1/ A, B, T, D, DIEL, LOSS, SIGM, P, XXP, YYP, XXPP, YYPP, L,
*          FO, NFRREQ, K0, TWOPI, PDO2A, PXPOA, BOA, BAND, PI,
*          ETAO, KK, GAIN, DELTAF
* IF (YYP .LT. YYPP) GO TO 10
* XP = XXP
* YP = YYPP
* XXP = XXP
* YYP = YYP
* GO TO 20
10 XP = XXP
YP = YYP
XXP = XXP
YPP = YYPP
20 TAU = 1.0
IF (YP .EQ. 0.) TAU = 2.
KSQ = DIEL * K0**2 * CMPLX(1.,-DELTA)
K = CSQRT(KSQ)
KODE = 1
Z12 = CCOS(K*(B-YPP)) * CCOS(K*YP) / (K*CSIN(K*B))
IF (ABS(YP - YPP) .LT. 0.001) KODE = 2
IF (KODE .EQ. 2) Z12 = Z12 + S3 * A
M = 0
SUBTOT = (0.,0.)
PXPOA = PI * XP / A
PXPOA = PI * XPP / A
1 M = M + 1
MPI = M * PI
PM = CSQRT(KSQ - (MPI/A)**2)
PMB = PM * B
PYP = PM * YP
PBMYPP = PM * (B - YPP)
TERM = 2*CCOS(PBMYPP) * CCOS(PYP) / (PM*CSIN(PMB))
IF (KODE .EQ. 2) TERM = TERM + TAU*A/MPI
TERM = TERM * COS(M*PXPOA) * COS(M*PXPOA) * J0(M*PDO2A)**2
SUBTOT = SUBTOT + TERM
IF (3*(M/3) .NE. M) GO TO 1
Z12 = Z12 + SUBTOT
IF (CABS(SUBTOT) .LT. 0.000001 * CABS(Z12)) GO TO 2
SUBTOT = (0.,0.)
GO TO 1
2 Z12 = -Z12 * CMPLX(0.,1.) * K0 * T * ETAO / A
CALL VZ1 (Z11, DELTA, XXP, YYP, S1)
CALL VZ1 (Z22, DELTA, XXPP, YYPP, S2)
RETURN
END

```

SUBROUTINE VZTOS(Z11,Z12,Z22,S11,S12,S22)

* PURPOSE: This program converts the open circuit parameters to
* scattering parameters refered to a 50 ohm system.
*
* PARAMETERS: The open circuit parameters are Z11, Z12, and Z22 and are
* are converted to the scattering parameters S11, S12, and S22.

COMPLEX Z11,Z12,Z22,S11,S12,S22
Z0=50
S11=((Z11-Z0)*(Z22+Z0)-Z12**2)/((Z11+Z0)*(Z22+Z0)-Z12**2)
S12=Z0*Z12/((Z11+Z0)*(Z22+Z0)-Z12**2)
S22=((Z11+Z0)*(Z22-Z0)-Z12**2)/((Z11+Z0)*(Z22+Z0)-Z12**2)
RETURN
END

INPUT/OUTPUT AND PLOTTING

N/S

```
DECODE (5,12,IYP) YP
IF (IXP .EQ. "      ") XP = XPO
IF (IYP .EQ. "      ") YP = YPO
XPO = XP
YPO = YP
GO TO 13
10 WRITE (1,11)
11 FORMAT (1*"*-* INPUT DATA IS BAD / JOB ABORTED")
STOP
12 FORMAT (F5.0)
13 WRITE (1,14)
14 FORMAT (^FO .^B^A^N^D. @B^/_^F .")
READ (2,15) FO,BAND,DELTAF
15 FORMAT (F8.0,X,F4.0,X,F4.0)
I = EOF(2)
IF (FO .EQ. 0) FO = FOO
IF (BAND .EQ. 0.) BAND = BANDO
IF (DELTAF .EQ. 0) DELTAF = DELTAO
IF (FO * BAND * DELTAF .EQ. 0) GO TO 10
BANDO = BAND
DELTAO = DELTAF
FOO = FO
NFREQ = 1
IF (DELTAF .NE. 0.) NFREQ = 0.5 + BAND*FO/(100*DELTAF)+1
IF (NFREQ .EQ. 2*(NFREQ/2)) NFREQ = NFREQ + 1
RETURN
END
```

SUBROUTINE VPLTPT (I2, F)

* PURPOSE: This subroutine plots the patterns of a rectangular microstrip antenna.

* PARAMETERS: I2 is the current calculated to flow through port 2
when a load impedance (found in VLOAD) is attached to it. F is the frequency.

* OPTIONS: Both polarizations in two planes, X-Z and Y-Z are plotted if the "linear option" is chosen. If the "CP" (circular polarization) option is chosen, then the response of a rotating dipole is simulated for the two aforementioned planes. All scales are linear (not dB). If the "individual normalization" option is chosen, then each pattern is normalized by its own maximum over the scan. The relative "cross-pol" cannot be seen with this option. At the end of each plot, a character must be input. The characters that are allowed have the following meanings:

CHARACTER	STANDS FOR	EFFECT
S	status	Next plot is same type as previous plot.
M	manual	Ask for options for next plot.
A	automatic	Same as "S" except no further input of options are possible and the plot is automatically copied.
R	re-plot	Re-plot last graph.
T	terminate	Terminate all plotting.

```

COMPLEX ET, ET1, EP, EP1, I2
INTEGER P, ANS, NORM, CP, FREQ(2)
COMMON /DELTA/ DELTA
COMMON /OPT/ ANS
COMMON /JID/ ID
COMMON /I1/ AZ, B, T, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,
*          FO, NFREQ, K0, TWOPI, P02A, PXPOA, BOA, BAND, PI,
*          ETAO, K, GAIN, DELTAF
* DIMENSION EXZX(721), EXZY(721), EYZX(721), EYZY(721), ID(3),
1          C(8), S(8)
DATA NORM /0/, INITIAL /0/
IF (INITIAL .NE. 0) GO TO 400
IXZ = "X-Z;"
IYZ = "Y-Z;"
IO = "O;"
IDASH = "-;"
ISLASH = "/";
PIBY180 = PI/180.
PIBY2 = PI/2.
PIBY8 = PI / 8
DO 300 J = 1, 8
ANGLE = (J-1) * PIBY8
C(J) = COS(ANGLE)
S(J) = SIN(ANGLE)
300 IF (NORM .NE. 0) GO TO 77
PRINT *, " "
PRINT *, " "

```

```

PRINT *, "*****"
PRINT *, "CP OR LINEAR (TYPE C OR L) ", N/S
READ 75, CP
IF (CP .EQ. "C") GO TO 76
FACTOR = PIBY180
FACT = 1.0
DENSE = 91
LIMIT = 181
PRINT *, "*****"
PRINT *, "INDIVIDUAL NORMALIZATION (Y OR N) ", N/S
READ 75, IQ
75 FORMAT (A1)
GO TO 77
76 FACTOR = PIBY180/4.
FACT = 0.25
DENSE = 361
LIMIT = 721
77 DO 3 J = 1, LIMIT
IF (CP .EQ. "C") IPSI = MOD(J-1,8) + 1
ANGLE = (DENSE-J) * FACTOR
CALL VPAT (ANGLE, 0., ET, EP, DELTA, (1.,0.), XP, YP)
IF (P .EQ. 1) GO TO 1
CALL VPAT (ANGLE, 0., ET1, EP1, DELTA, I2, XPP, YPP)
EP = EP + EP1
ET = ET + ET1
1 IF (CP .EQ. "C") GO TO 111
EXZX(J) = CABS(ET)
EXZY(J) = CABS(EP)
GO TO 112
111 EXZX(J) = CABS(EP*C(IPSI)+ET*S(IPSI))
112 CALL VPAT (ANGLE, PIBY2, ET, EP, DELTA, (1.,0.), XP, YP)
IF (P .EQ. 1) GO TO 2
CALL VPAT (ANGLE, PIBY2, ET1, EP1, DELTA, I2, XPP, YPP)
EP = EP + EP1
ET = ET + ET1
2 IF (CP .EQ. "C") GO TO 22
EYZX(J) = CABS(EP)
EYZY(J) = CABS(ET)
GO TO 3
22 EYZY(J) = CABS(EP*C(IPSI) + ET*S(IPSI))
IPSI = MOD(J-1,8) + 1
3 CONTINUE
CALL URESET
A1 = 0.
A2 = 0.
A3 = 0.
A4 = 0.
DO 5 J = 1, LIMIT
A1 = AMAX1(A1, EXZX(J))
IF (CP .NE. "C") A2 = AMAX1(A2, EXZY(J))
IF (CP .NE. "C") A3 = AMAX1(A3, EYZX(J))
5 A4 = AMAX1(A4, EYZY(J))
A = AMAX1 (A1, A2, A3, A4)
CALL UPEN (0., 0.)
7 CALL UERASE
CALL UDAREA (0.4, 5.119, 0.4, 5.119)
CALL UWINDO (-1., 1., -1., 1.)
DO 8 J = 1, 5
8 CALL UCRCLE (0., 0., 0.2 * J)
CALL USET ("POLAR")
DO 9 J = 1, 36
ANGLE = J * 10
CALL UMOVE (0.2, ANGLE)
9 CALL UPEN (1., ANGLE)
DO 10 J = 1, 4
ANGLE = J * 90
CALL UMOVE (0., 0.)
10 CALL UPEN (0.2, ANGLE)
CALL USET ("LINE")

```

```

AA = A
IF (IQ .EQ. "Y") AA = A1
CALL UMOVE (EXZX(1)/AA,90.)
DO 11 J = 1, LIMIT
ANGLE = (DENSE - J) * FACT
11 CALL UPEN (EXZX(J)/AA, ANGLE)
IF (IQ .EQ. "Y") AA = A3
CALL UMOVE (EYZY(1)/AA,90.)
DO 12 J = 1, LIMIT
ANGLE = (J + DENSE - 2) * FACT
12 CALL UPEN (EYZY(J)/AA, ANGLE)
IF (IQ .EQ. "Y") AA = A2
IF (CP .EQ. "C") GO TO 120
CALL USET ("RECT")
CALL UMOVE (0.8, -0.8)
CALL UPEN (1.0, -0.8)
CALL USET ("POLAR")
CALL USET ("DASHLINE")
CALL UPSET ("SETDASH", 5212.)
CALL UMOVE (EXZY(1)/AA,90.)
DO 13 J = 1, 181
ANGLE = (91 - J)
13 CALL UPEN (EXZY(J)/AA, ANGLE)
IF (IQ .EQ. "Y") AA = A4
CALL UMOVE (EYZX(1)/AA,90.)
DO 14 J = 1, 181
ANGLE = (J + 89)
14 CALL UPEN (EYZX(J)/AA, ANGLE)
CALL USET ("RECT")
CALL UMOVE (0.8, -0.9)
CALL UPEN (1.0, -0.9)
CALL UDAREA (0., 6., 0., 5.119)
CALL UPRINT (0.76, -0.688, IO)
CALL UPRINT (0.76, -0.768, IO)
CALL UPRINT (0.76, -0.688, IDASH)
CALL UPRINT (0.76, -0.768, ISLASH)
GO TO 121
120 CALL USET ("TEXT")
CALL USET ("RECT")
CALL UDAREA (0., 6., 0., 5.119)
121 CALL UPRINT (-1.0, 0., IYZ)
CALL UPRINT (+0.76, 0., IXZ)
CALL UPRINT (-0.40, -1.0, ID)
CALL USET ("LINE")
ENCODE (11,16 FREQ) F
16 FORMAT ("F = ",F6.1,".")
CALL UPRINT (0.60,-1.0,FREQ)
IF (NORM .NE. 2) GO TO 15
CALL UFLUSH
PRINT # "^\^&
CALL UPause
RETURN
15 CALL UREAD (-1., -1., KQ, 1., FLAG)
IF (KQ .EQ. "S") NORM = 1
IF (KQ .EQ. "M") NORM = 0
IF (KQ .EQ. "A") NORM = 2
IF (KQ .EQ. "R") GO TO 7
IF (KQ .EQ. "T") ANS = 3
CALL UERASE
RETURN
END

```

N/S

N/S

SUBROUTINE VPLTS (F, Z11, Z12, Z22, IACC)

```

* PURPOSE: This routine plots the "s" parameters of a two port
* microstrip antenna using the input values of the
* open circuit parameters. It uses the software of the GCS
* system.

* PARAMETERS: F is the frequency in MHz.
* Z11, Z12, and Z22 are the respective open circuit
* parameters of the microstrip antenna (input).
* IACC is set equal to zero when data is being
* accumulated by the subroutine for later plotting.
* It is set to unity when the accumulated data is
* to actually be plotted.

* OPTIONS: At the end of each plot, one may input either a blank
* character or an "R". The latter causes the system to
* replot the graph.

*****COMPLEX S11(100), S12(100), S22(100), Z11, Z12, Z22
DIMENSION ID(3)
INTEGER ONEONE, ONETWO, TWOTWO, S
COMMON /JID/ ID
COMMON /I1/ A, B, T, D, DIEL, LOSS, SIGM, P, XP, YP, XPP, YPP, L,
* F0, NFREQ, K0, TWOP1, PDO2A, TPPOA, BOA, BAND, PI,
* ETAO, K, GAIN, DELTAF
COMMON /DELTA/ DELTA
S = "S:"
IX = "X,"
IO = "O,"
IS = "+,"
IC = "#,"
ONEONE = " 11;"
ONETWO = " 12;"
TWOTWO = " 22;"
ID(3) = ";"
IF (IACC .EQ. 1) GO TO 5
CALL VZTOS (Z11, Z12, Z22, S11(K), S12(K), S22(K))
IF (K .GT. 1) GO TO 3
Q = 1 / DELTA
CALL DATE (ID(1))
CALL TIME (ID(2))
WRITE (3,1) A, B, T, D, XP, YP, XPP, YPP, DIEL, LOSS,
* SIGM, F0, DELTAF, Q, GAIN, ID(1), ID(2)
1 FORMAT("1",6X,"T W O P O R T R E C T A N G U L A R M I C R O
1 S T R I P /",7X,"_
2 - - - - - /",7X,"Dimensions (a X b X t)
3 (",F4.1," X ",F4.1," X ",F4.2,") cm",/,7X,"F"
4 ed data: (1) Width ..... (",F4.1," cm",/,7X,"F"
5 ",/,18X,"(2) Port one ..... (",F4.1," ",F4.1," "
6 ) cm",/,18X,"(3) Port two ..... (",F4.1," ",F4.1," "
7 F4.1,") cm",/,7X,"Electrical parameters: (1)
8 Dielectric ..... ",F5.2,"/30X,"(2) Loss tangent
9 ",F7.5,"/30X,"(3) Conductivity ... ",F5.1," KMho
* s/Cm",/,7X,"Center frequency ..... "
1..... ",F7.1," MHz",/,7X,"Frequency increment
2..... ",F7.1," MHz",/,7X,"Quality fact
3 or ..... ",F5.1,"/7X,"Gain .....
4 .. ",F4.1," db",/,7X,"Date/time .....
5 .. ",F5.1,"/7X,"Time .....
6 Y",11X,"S",16X,"S",16X,"S",/9X,"(MHz)",14X,"11",15X,"12",15X,"12
7 ",/ "0")
3 WRITE (3,4) F, S11(K), S12(K), S22(K)
4 FORMAT (7X,F7.1,6X,3("(",F6.3,",",F6.3,")",2X))
RETURN

```

```

5 CALL URESET
CALL UERASE
NO2 = NFREQ/2 + 1
NO21 = NO2 + 1
CALL USET ("LINE")
CALL UWINDO (-1., 1., -1., 1.)
CALL UDAREA (0.4, 5.119, 0.4, 5.119)
CALL UCRCLE (0., 0., 1.)
CALL UMOVE (0., 1.)
CALL UARC (1., 1., 90.)
CALL UARCL (1., -1., 90.)
CALL UCRCL (0.5, 0.0, 0.5)
CALL UMOVE (-1.0, 0.0)
CALL UPEN (1.0, 0.0)
CALL USET ("DASHLINE")
CALL UPSET("SETDASH",5414.)
CALL UCRCLE (0., 0., 0.5)
CALL USET("LINE")
CALL USET ("NX")
CALL USET ("SOFT")
CALL UPSET ("HORIZONTAL", 0.03)
CALL UPSET ("VERTICAL", 0.03)
DO 6 J = 1, NFREQ
IF (J .EQ. NO2) CALL USET ("N*")
IF (J .EQ. NO21) CALL USET ("NX")
6 CALL UPEN (REAL(S11(J)), AIMAG(S11(J)))
CALL UPRINT (0.75, -0.75, IX)
CALL USET ("NO")
DO 7 J = 1, NFREQ
IF (J .EQ. NO2) CALL USET ("N*")
IF (J .EQ. NO21) CALL USET ("NO")
7 CALL UPEN (REAL(S12(J)), AIMAG(S12(J)))
CALL UPRINT (0.75, -0.85, IO)
CALL USET ("N+")
CALL UPSET ("VERTICAL", 0.06)
CALL UPSET ("HORIZONTAL", 0.04)
DO 8 J = 1, NFREQ
IF (J .EQ. NO2) CALL USET ("N*")
IF (J .EQ. NO21) CALL USET ("N+")
8 CALL UPEN (REAL(S22(J)), AIMAG(S22(J)))
CALL UPRINT (0.75, -0.95, IS)
CALL USET ("HARD")
CALL UPRINT (0.85, -0.80, ONEONE)
CALL UPRINT (0.85, -0.775, S)
CALL UPRINT (0.85, -0.975, S)
CALL UPRINT (0.85, -0.875, S)
CALL UPRINT (0.85, -0.90, ONETWO)
CALL UPRINT (0.85, -1., TWOTWO)
CALL UDAREA (0., 5.5, 0., 5.5)
CALL UPRINT (-0.47, -1.0, ID)
CALL UREAD (-1., -1., IQ, 1., F)
IF (IQ .NE. " ") GO TO 5
CALL UERASE
RETURN
END

```

```

SUBROUTINE VPLTZ (Z, F, IACC)
* PURPOSE: This program accumulates impedance data, outputs in tab-
*  ulated form, and then plots in one of three optional ways:
*   (a) Smith chart, (b) Rectangular G-B plot, or (3) magnitud
*   of reflection coefficient vs frequency.
*
* PARAMETERS: Z is the complex impedance to be plotted.
*   F is the frequency at which Z was determined.
*   IACC is 2, data is accumulated for future plotting,
*   while if IACC is 1, the data is plotted.

COMPLEX Y, GAMMA, Z
INTEGER P, STRING(4)
DIMENSION ID(3),FREQ(100),G(100),B(100),IF0(2),IINC(2),IDIM(2)
COMMON /JID/ ID
COMMON /LDID/ STRING
COMMON /I1/ A,BB,T,D,DIEL,LOSS,SIGM,P,XP,YP,XPP,YPP,L,
*           F0,NFREQ,K0,TWOP1,PDO2A,TPXPOA,BOA,BAND,Pi,
*           ETAO,K,GAIN,DELTAf
COMMON /DELT/ DELTA
ID(3) = ";"
IF (IACC .EQ. 1) GO TO 10
N02 = NFREQ/2 + 1
N021 = N02 + 1
IF (K .GT. 1) GO TO 3
Q = 1 / DELTA
CALL DATE (ID(1))
CALL TIME (ID(2))
ENCODE (17, 202, IDIM) A, BB
202 FORMAT {F5.2,"X",F5.2," CM;"}
ENCODE (17 200, IF0) F0
200 FORMAT {"F0 = ",F6.1," MHZ;"}
ENCODE (16 201, IINC) DELTAf
201 FORMAT {"INC = ",F5.1," MHZ;"}
WRITE (3,1) A, BB, T, D, XP, YP
IF (P .EQ. 2) WRITE (3,999) XPP, STRING
WRITE (3,1000) DIEL, LOSS, SIGM, F0, DELTAf, Q, GAIN, ID(1), ID(2)
1 FORMAT({"1",6X,"T W O P O R T R E C T A N G U L A R M I C R O
1 S T R I P",/7X,"-----,
2 -----,"0",6X,"Dimensions (a X b X t) .....
3 .....,("F5.2," X ",F5.2," X ",F4.2,") cm",/,7X,"F"
4 ed data: (i) Width ..... ("F4.2," cm
5 ",/,18X,"(2) Port one ..... ("F5.2," ",F5.2,
6 ) cm")
999 FORMAT(18X,"(3) Port two ..... (",
1F5.2," ",F5.2,) Cm",/,
27X,"Load description: ..... ",
34A10)
1000 FORMAT(
8Dielectric ... ,F5.2,/30X,"(2) Loss tangent
9 .....,F7.5./30X,"(3) Conductivity ... ",F5.1," KMoh
8/Cm",/7X,"Center frequency .....
1..... ,F7.1," MHz",/7X,"Frequency increment
2..... ,F7.1," MHz",/7X,"Quality fact
3or .....,F5.1,/7X,"Gain .....
4..... ,F4.1," dB",/7X,"Date/time ..
5..... ,F2A10/,0",6X,"Frequency
6Y",1jX,"Z",/9X,"(MHz)",/
7, "0")
3 WRITE (3,4) F, Z
4 FORMAT (7X,F7.1,6X,"(",F6.2,",",F7.2,")")
FREQ(K) = F - F0
Y = 50./Z
G(K) = REAL(Y)

```

```

B(K) = AIMAG(Y)
RETURN
10 CALL URESET
PRINT *, "*****"
PRINT *, "*****"
PRINT *, "*****"
PRINT *, "*****"
PRINT *, "RECTANGULAR, SMITH CHART, OR VS FREQUENCY PLOT (R, S, ",
1     "OR V)",
READ 2, KODE
2 FORMAT (A1)
IF (EOF(2) .NE. 0) RETURN
IF (KODE .EQ. "S") KKODE = 1
IF (KODE .EQ. "R") KKODE = 2
IF (KODE .EQ. "V") KKODE = 3
GO TO (5, 12, 15), KKODE
5 CALL UERASE
CALL USET ("LINE")
CALL UWINDO (-1., 1., -1., 1.)
CALL UDAREA {0.4, 5.119, 0.4, 5.119}
CALL UCRCLE {0., 0., 1.}
CALL UMOVE (0., 1.)
CALL UARC {1., 1., 90.}
CALL UARC {1., -1., 90.}
CALL UCRCLE {0.5, 0.0, 0.5}
CALL UMOVE (-1.0, 0.0)
CALL UPEN (1.0, 0.0)
CALL USET("DASHLINE")
CALL UPSET("SETDASH", 5414.)
CALL UCRCLE {0., 0., 0.5}
CALL USET("LINE")
CALL USET ("NX")
CALL USET ("SOFT")
CALL UPSET ("HORIZONTAL", 0.02)
CALL UPSET ("VERTICAL", 0.02)
DO 6 J = 1, NFREQ
Y = CMPLX {G(J), B(J)}
GAMMA = (1-Y)/(1+Y)
IF (J .EQ. NO2) CALL USET ("N*")
IF (J .EQ. NO21) CALL USET ("NX")
6 CALL UPEN (REAL(GAMMA), AIMAG(GAMMA))
CALL USET ("HARD")
CALL UDAREA {0., 5.5, 0., 5.5}
CALL UPRINT {-0.40, -1.0, ID}
CALL USET ("SOFT")
CALL USET ("ITALICS")
CALL UPSET ("HORIZONTAL", 0.03)
CALL UPSET ("VERTICAL", 0.05)
CALL UPRINT {0.4, -0.85, IFO}
CALL UPRINT {0.4, -0.92, IINC}
CALL UPRINT {0.4, -0.99, IDIM}
CALL UERASE
14 CALL UREAD (0., 0., KODE, 1., FLAG)
IF (KODE .EQ. "R") GO TO 10
GO TO 9
12 RMAX = 0.
XABSMX = 0.
DO 13 J = 1, NFREQ
Z = 1 / CMPLX {G(J), B(J)}
XABSMX = AMAX1 (ABS(AIMAG(Z)), XABSMX)
13 RMAX = AMAX1 (RMAX, REAL(Z))
ABSMAX = AMAX1 (RMAX, XABSMX)
DEC = ALOG10 (ABSMAX)
IDEC = DEC
IF (DEC .LT. 0.) IDEC = IDEC - 1
DECADE = 10. ** IDEC
IDIGIT = ABSMAX / DECADE + 1
WINDOW = IDIGIT * DECADE
CALL UWINDO (0., 2.*WINDOW, -WINDOW, WINDOW)

```

```

TICK = DECADE
IF (IDIGIT .LT. 7) TICK = 0.5 * DECADE
IF (IDIGIT .LT. 3) TICK = 0.2 * DECADE
CALL UPSET ("TICX", TICK)
CALL UPSET ("TICY", TICK)
CALL UERASE
CALL UDAREA (0.4, 5., 0.4, 5.)
CALL UPSET ("YLABEL", "X;")
CALL UPSET ("XLABEL", "R;")
CALL USET ("GRIDAXIS")
CALL USET ("XBOTH")
CALL USET ("YBOTH")
CALL USET ("POINT")
CALL USET ("OWNSCALE")
CALL UAXIS (0., 2.*WINDOW, -WINDOW, WINDOW)
CALL USET ("LINE")
CALL USET ("NX")
CALL USET ("SOFT")
CALL UPSET ("HORIZONTAL", 0.02 * WINDOW)
CALL UPSET ("VERTICAL", 0.02 * WINDOW)
DO 130 J = 1, NFREQ
Z = 1 / CMPLX (G(J), B(J))
130 CALL UPEN (REAL(Z), AIMAG(Z))
CALL USET ("HARD")
CALL UDAREA (0., 5., 0., 5.)
CALL UWINDO (-1., 1., -1., 1.)
CALL UPRINT (-0.47, -1.0, ID)
CALL UPAUSE
CALL UERASE
CALL USET ("LINE")
GO TO 14
15 CALL UERASE
CALL UDAREA (0.4, 5.4, 0.4, 5.4)
CALL USET ("GRIDAXES")
CALL USET ("XBOTH")
CALL USET ("YBOTH")
CALL UPSET ("XLABEL", "FREQUENCY - F0;")
CALL UPSET ("YLABEL", "MAGNITUDE OF GAMMA;")
CALL USET ("AUTO")
CALL UAXIS (FREQ(1), FREQ(NFREQ), 0.00, 1.00)
CALL USET ("SOFT")
CALL UPSET ("HORIZONTAL", 0.02 * (FREQ(NFREQ) - FREQ(1)))
CALL UPSET ("VERTICAL", 0.02)
CALL USET ("NX")
DO 156 J = 1, NFREQ
Y = CMPLX (G(J), B(J))
GAMMA = (1-Y)/(1+Y)
156 CALL UPEN (FREQ(J), CABS(GAMMA))
CALL USET ("SOFT")
CALL USET ("ITALLICS")
CALL UDAREA (0., 7.49, 0., 5.71)
CALL UWINDO (0., 1., 0., 1.)
CALL UPSET ("HORIZONTAL", 0.015)
CALL UPSET ("VERTICAL", 0.03)
CALL UPRINT (0.7, 0.5, ID)
CALL UPRINT (0.7, 0.47, IFO)
CALL UPRINT (0.7, 0.44, IINC)
CALL UPRINT (0.7, 0.41, IDIM)
CALL UPAUSE
CALL UERASE
GO TO 14
9 RETURN
END

```

CHAPTER 3: EXAMPLES

This chapter contains examples of the use of this program to analyze a microstrip antenna. The specific case chosen was that of a nearly square microstrip antenna. This was chosen to illustrate that by making one side of the antenna a small amount larger than the other, and by properly feeding and loading the the antenna with a variable capacitor (such as a varactor), the antenna can be switched from left hand circular polarization to right hand circular polarization. This is consistent with both theory and experiments carried out at the University of Illinois and presented in references [4] and [6]. The output listed below is a copy of the actual graphical data displayed on a graphics terminal by the program. Explanatory remarks have been added to aid the reader.

```

? a 7.62 b 7.80 t .15 d.diel.loss.sigm.p.
? x' y'
? f0 3.90
? band. af .
? 1216.0 10. 2.

```

Choose an option:

- (1) Plot patterns at all frequencies.
 - (2) Plot pattern only at center frequency.
 - (3) Plot no patterns
- Type option (1, 2, or 3):

? 2

CP OR LINEAR (TYPE C OR L) ? L

INDIVIDUAL NORMALIZATION (Y OR N) ? N

The option of plotting a pattern at the center frequency was chosen in this case. Additionally, within the plotting routine itself, the options of a linear polarization pattern without individual normalization were chosen.

In this example, a , b , and t are the geometrical parameters (in cm) of the microstrip antenna shown in Fig. 2. The parameter d is the "effective feed width" in cm. The dielectric constant relative to free space is the input under the "diel." The loss tangent of the dielectric (times one thousand) is typed under the "loss." The conductivity of the cladding is under the "sigm" in units of K/cm. The parameter, p , is the number of ports (either "1" or "2"). The default value of p is initially unity and its previous value in subsequent computations. The coordinate of the feed point is (x', y') corresponding to (x_1, y_1) in the discussion in Appendix 1. The impedance, and, optionally, the patterns will be computed for frequencies, f , in a range of approximately

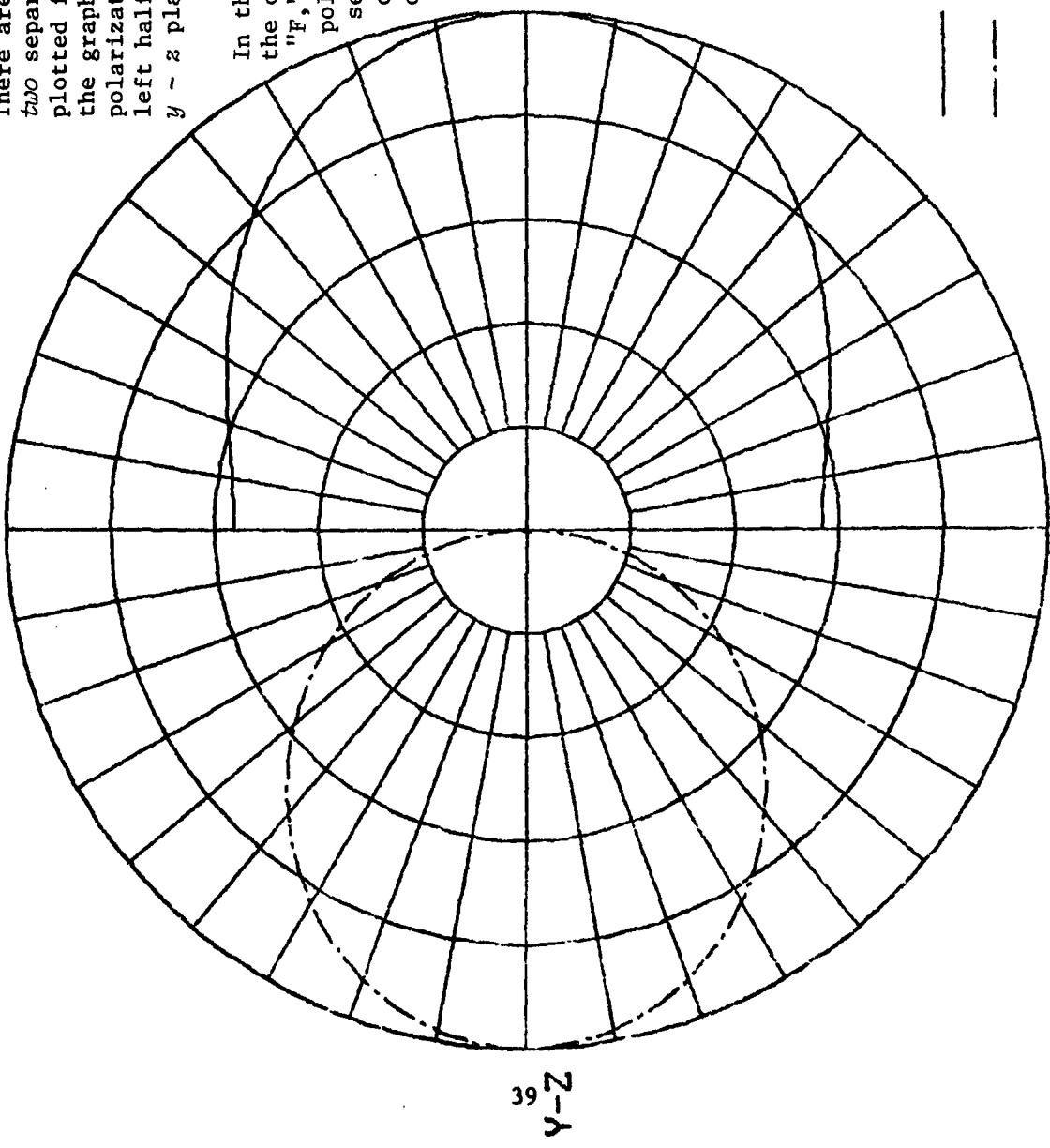
$$\left[1 + \frac{\eta}{200}\right] f_0 \leq f \leq \left[1 - \frac{\eta}{200}\right] f_0$$

where η is typed in under "band" (in percentage), and f_0 is typed in under "f0." Computations are always performed at f_0 and at frequencies differing from f_0 within the band by integral multiples of Δf , the input under " Δf ."

There are two separate patterns in each of two separate planes. Both E_0 and E_ϕ are plotted for each plane. The right half of the graph represents a plot of these two polarizations in the $x - z$ plane, while the left half is a plot of these patterns in the $y - z$ plane.

In this particular example, because of the choice of feed point and frequency, "F," printed at the bottom, the cross-polarization is so small it cannot be seen. For the case where "cross-pol" can be seen clearly, refer to the end of the example.

The group of numbers directly under the graph is the "date/time-group." This is used to identify the X-Z plot with the associated "hard copy" output that is automatically printed. It appears on all plots.



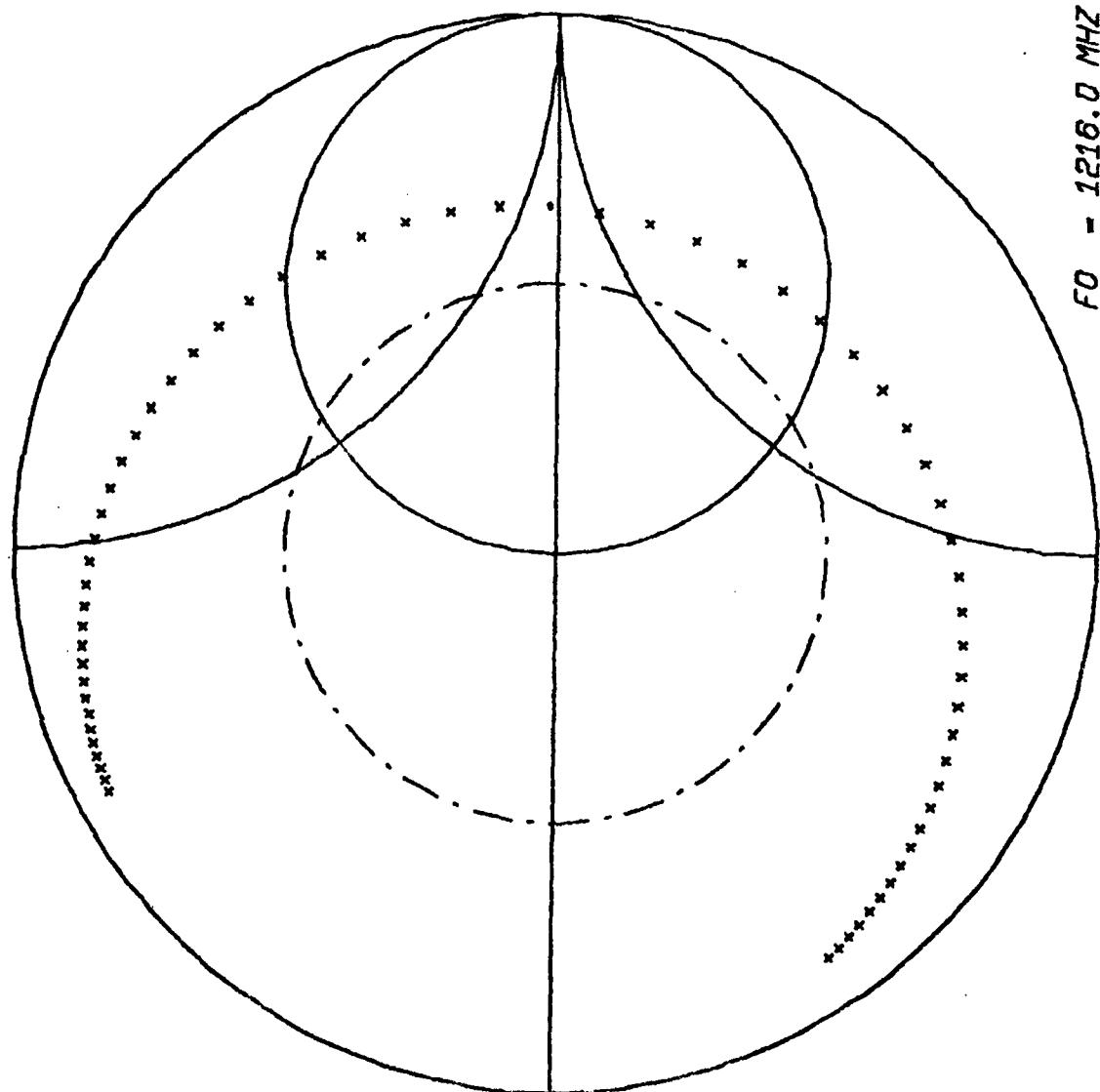
80/03/11. 11.20.42.

F = 1216.0

RECTANGULAR, SMITH CHART, OR VS FREQUENCY PLOT (R, S, OR V)? S

Here, the system is requesting the input of the desired type of impedance plot. If no plot is desired at all, the user can simply respond with a carriage return. In this case, the user chose a Smith chart plot.

This is a plot of the complex reflection coefficient. That is it is the impedance locus plotted on a Smith chart. The center frequency, $f_0 = F_0$, the increment in frequency between adjacent points in the Smith chart, $\Delta f = INC$, and the patch dimensions, $a \times b$, are printed with the plot. The asterisk, "*", is the point corresponding to f_0 .



$F_0 = 1216.0 \text{ MHz}$
 $INC = 2.0 \text{ MHz}$
80/03/11. 11.20.42. 7.62 X 7.80 CM

Continue (type Y or N)

42
? Y

```
? a . b . t . d . diel . loss . sigm . p .
? x' . y' .
? 3.81 . 1.00
? f0 . band . sf .
? 1188.0 10. 2.
Choose an option:
(1) Plot patterns at all frequencies.
(2) Plot pattern only at center frequency.
(3) Plot no patterns
Type option (1, 2, or 3):
? 2
```

In this implementation, only those parameters which are to be changed from the previous calculation need be input.

```
? CP OR LINEAR (TYPE C OR L) ? L
```

```
INDIVIDUAL NORMALIZATION (Y OR N) ? N
```

"Individual normalization," as used here means that all patterns are normalized by their own respective maximum values. Thus, information about the relative importance of cross polarization is lost under this option. The usual response is "N" (for "no"), the default. Under the "N" option, the patterns are all normalized by the same factor. This factor is the maximum of all four patterns.

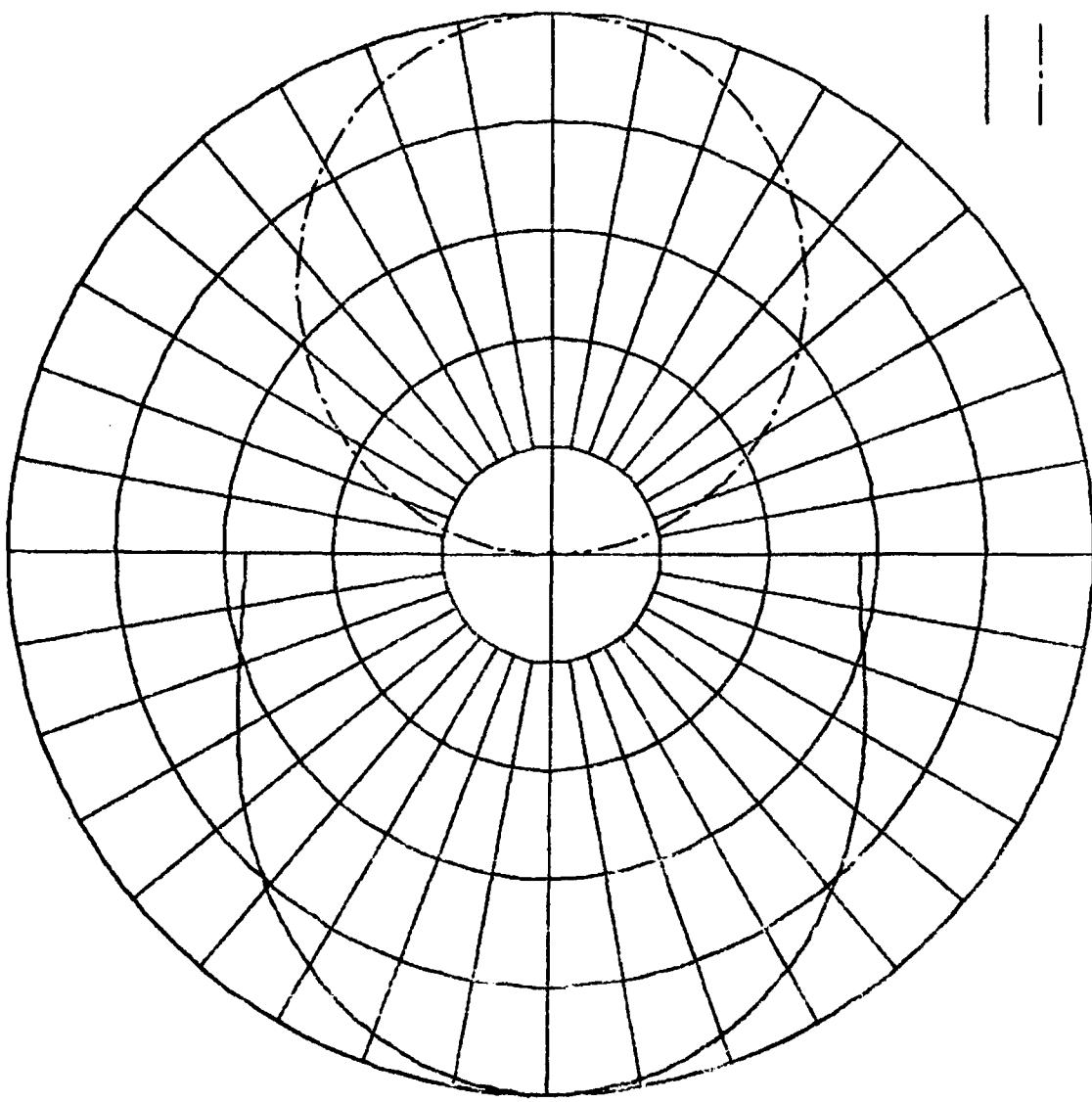
$F = 1188.0$

80/03/11. 11.26.31.

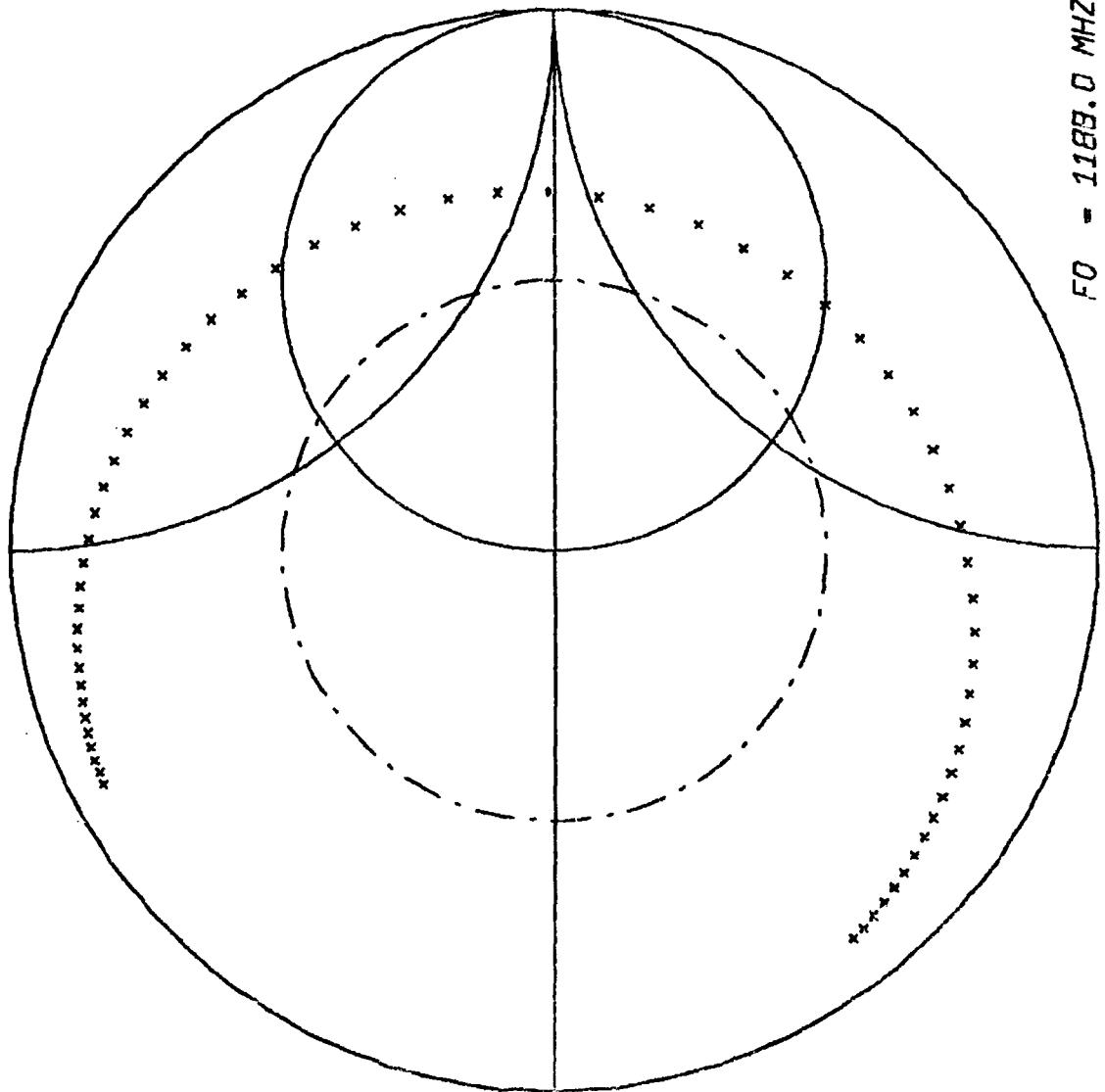
θ
 ϕ

X-Z

Y-Z
44



RECTANGULAR, SMITH CHART, OR VS FREQUENCY PLOT (R, S, OR V)? S



$F_0 = 1189.0 \text{ MHz}$
 $T_{NC} = 2.0 \text{ MHz}$
80/03/11. 11.26.31. 7.62 X 7.80 CM

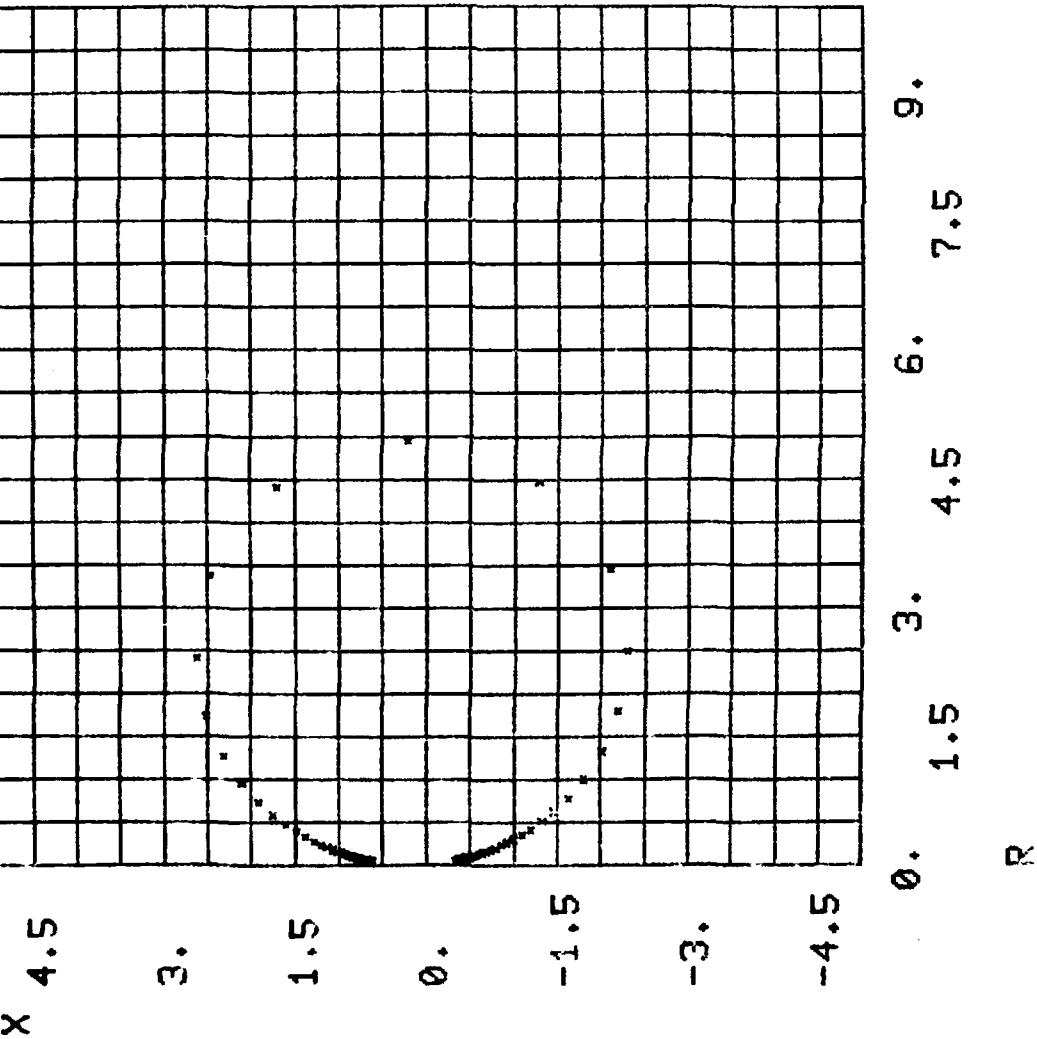
When the computer has completed an impedance plot, the program awaits (with no prompting) the input of a single character from the keyboard. If that character is an "R" (for "replot"), the program will again ask for a choice of the type of impedance plot desired and replot the same data. In this case, an "R" was input which caused the message below to be typed.

R

RECTANGULAR, SMITH CHART, OR VS FREQUENCY PLOT (R, S, OR U)? R

This time the user chose to plot in a rectangular system.

This type of graph plots the complex impedance in the impedance, Z , plane with the horizontal axis being $\text{Re } Z$ and the vertical axis $\text{Im } Z$.



80/03/11. 11.26.31.

Again, a replot was requested by entering an "R" from the keyboard.

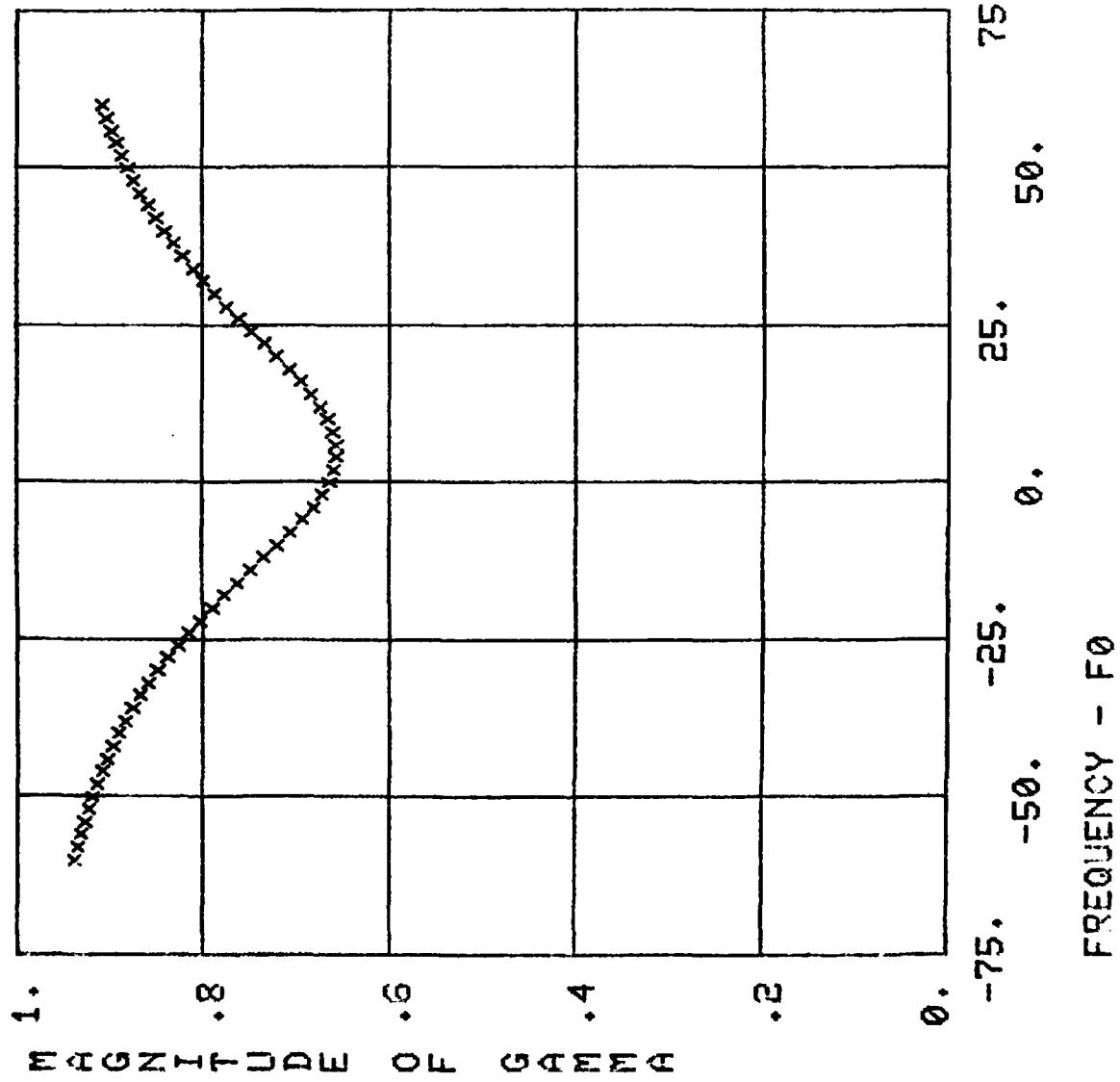
R

RECTANGULAR, SMITH CHART, OR VS FREQUENCY PLOT (R, S, OR V)? V

This time a "vs frequency" plot was requested.

This plots the magnitude of
reflection coefficient against
 $f - f_0$.

BO/03/11. 11.26.31.
 $F_0 = 1188.0 \text{ MHz}$
 $\text{INC} = 2.0 \text{ MHz}$
 $7.62 \times 7.80 \text{ cm}$



Continue (type Y or N)

? N

```
? 7.62 . b . t . d . die1 . loss . sigm . g .
? x , y , x , y " . L .
? 1.00 1.00 1.00 3.90
? f0 . band . sf .
? 1210.10.2.
```

choose an option:

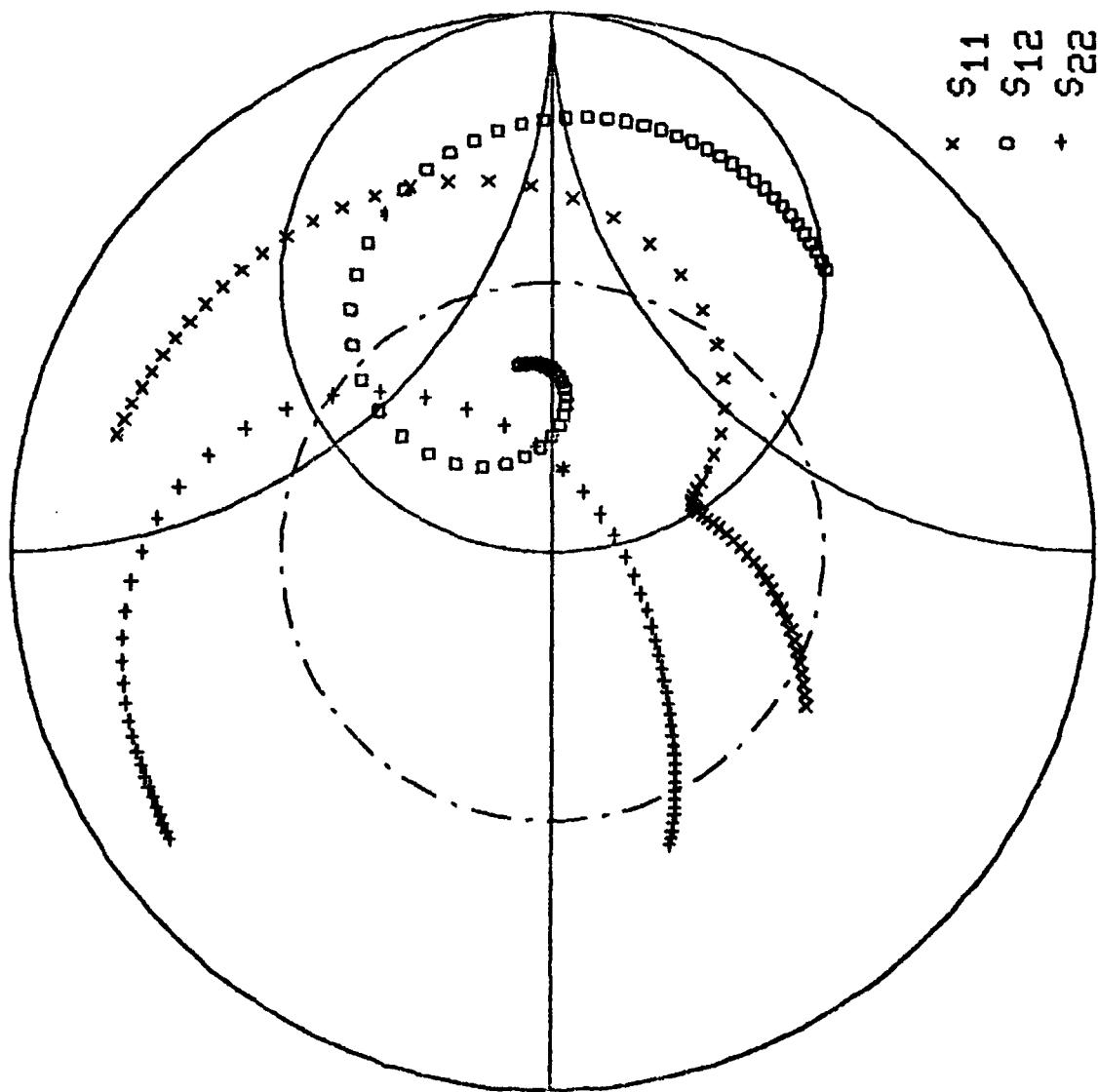
- (1) Plot patterns at all frequencies.
 - (2) Plot pattern only at center frequency.
 - (3) Plot no patterns
- Type option (1, 2, or 3):

? 3

In this case, a second port is added to the existing antenna. This is indicated by placing a "2" under the "P." The coordinate of the second port is (x^2, y^2) . This corresponds to (x_2, y_2) in the discussion in Appendix I. The "L" parameter allows the user to indicate whether or not there is a load on the second port by placing a "y" or "N" under it, respectively. The "N" option is assumed initially as default. Thereafter, if L is not specified, it is taken to be its previous value.

(Also note that since this was the beginning of a new run (the previous run having been terminated), all parameters had to be input.)

This is a Smith chart plot of the s-parameters of the unloaded two-port. Again, the location of points corresponding to the center frequency are indicated by "*"s.



80/03/11. 11.49.40.

Continue (type Y or N)

```
? a . b . t . d . diel . loss . sigm . p .
? x' . y' . x" . y" . L .
? f0 . band . sf .
? 1194 . 4 . 1.
```

Choose an option:

- (1) Plot patterns at all frequencies.
 - (2) Plot pattern only at center frequency.
 - (3) Plot no patterns
- Type option (1, 2, or 3):

```
? 2 INPUT CAPACITANCE IN PICOFARADS: C = ? 1.35
```

Here, a load was added to the second port as indicated by the "Y" below the "L." The type of load is determined by the user supplied subroutine, VLOAD, which must be loaded with the rest of the program prior to execution.

The particular load used in this example was the capacitor. The subroutine VLOAD was programmed to ask the user to input the capacitance, 1.35 pF in this case.

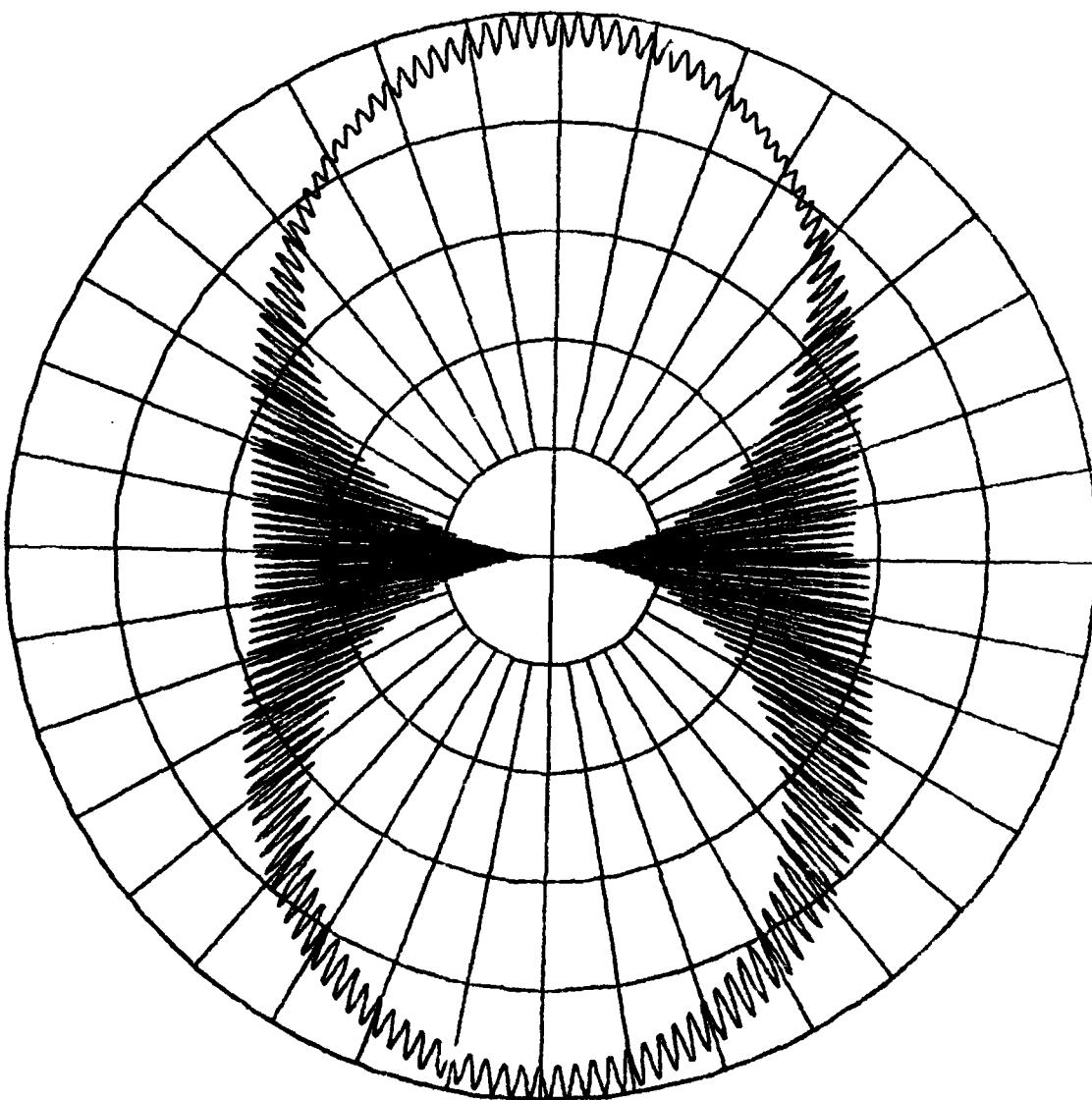
A plot at the center frequency for this example was also requested with the "C" option.

55 CP OR LINEAR (TYPE C OR L) ? C

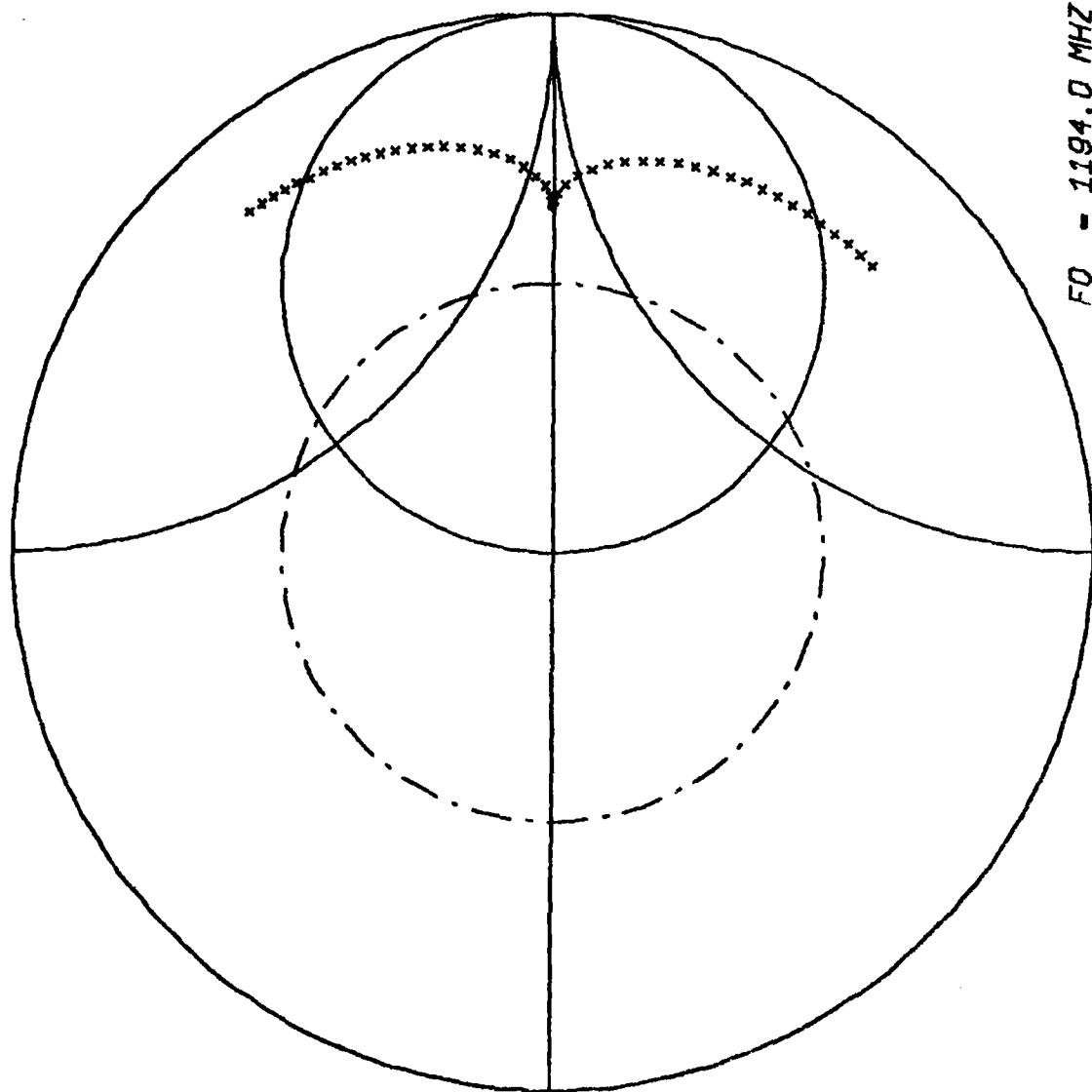
These patterns are the simulated responses of a rotating dipole which makes five full rotations per twenty degrees change in elevation angle, θ . Thus, when the antenna is producing circular polarization (CP) in a given direction, the oscillations in the response in that direction disappear.

In this plot, as in the linear polarization plots, the right half of the graph is the pattern in the $x - z$ Plane, while the left half is the pattern in the $y - z$ plane.

X-Z
 $y - z$ plane.



RECTANGULAR, SMITH CHART, OR VS FREQUENCY PLOT (R , S , OR V)? S



$F_D = 1194.0 \text{ MHz}$
 $I_{NC} = 1.0 \text{ MHz}$
80/03/11. 11.53.52. 7.62 X 7.80 CM

? . a . b . t . d .diel.loss.sigm.p.
? . x' . y' . x" . y" .L.
? . f0 .band. sf .
? 1182.0

Choose an option:

- (1) Plot patterns at all frequencies.
 - (2) Plot pattern only at center frequency.
 - (3) Plot no patterns
- Type option (1, 2, or 3):

? 2 INPUT CAPACITANCE IN PICOFARADS: C = ? 3.1

In this case, the value of the capacitance has been changed from 1.35 pF to 3.1 pF. This change in load has the effect of reversing the sense of CP.

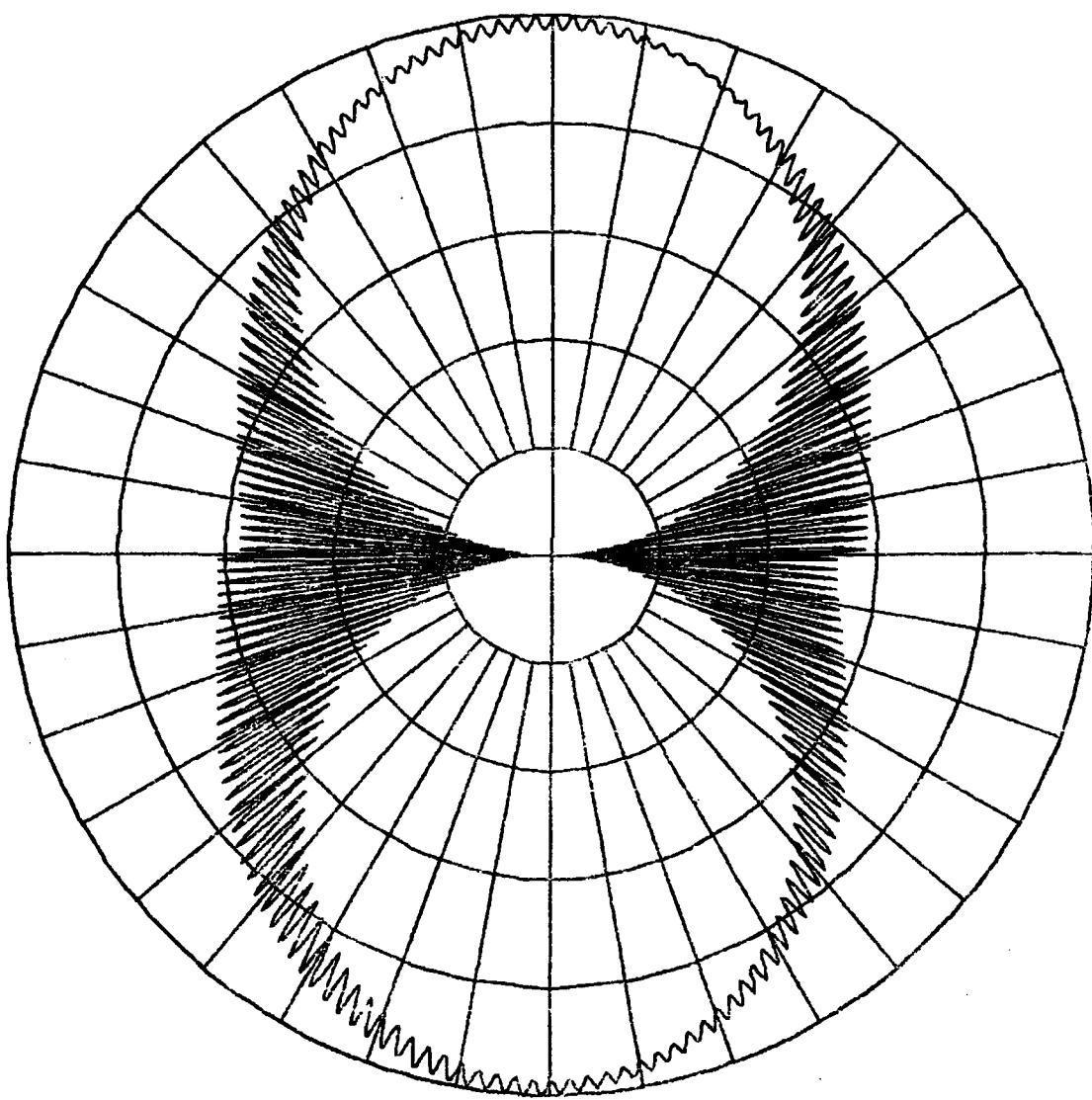
CP OR LINEAR (TYPE C OR L) ? C

F = 1182.0

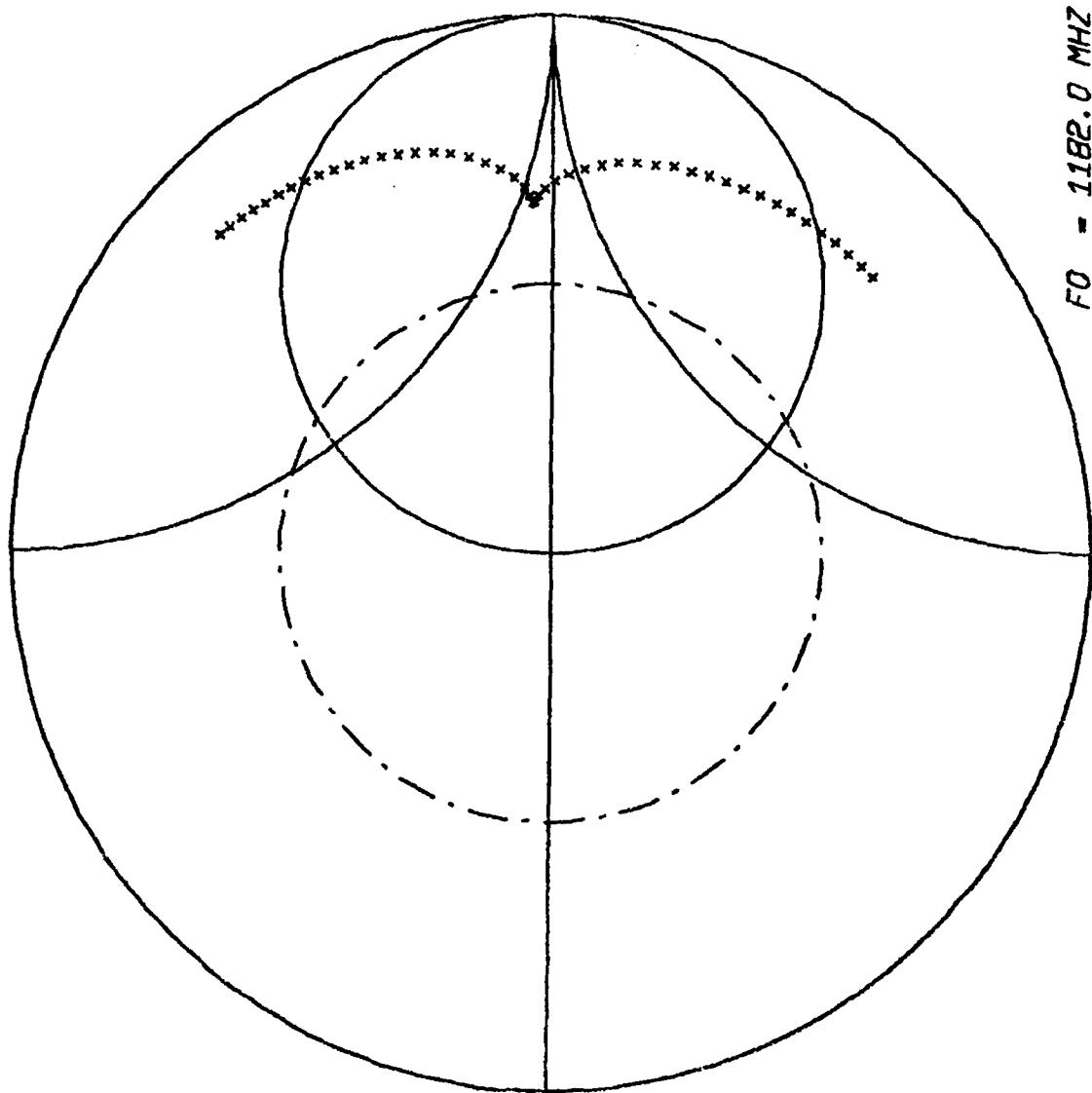
80/03/11. 11.59.54.

X-Z

Y-Z



RECTANGULAR, SMITH CHART, OR VS FREQUENCY PLOT (R, S, OR V)? S



$F_0 = 1182.0 \text{ MHz}$
 $\Delta\text{INC} = 1.0 \text{ MHz}$
80/03/11. 11.59.54. 7.62 X 7.80 CM

Continue (type Y or N)

? Y

? . a . b . t . d .diel.loss.sigm.p.
? . x' . y' . x" . y" .L.
? . f0 .band. af .
? .2 1.

Choose an option:

- (1) Plot patterns at all frequencies.
 - (2) Plot pattern only at center frequency.
 - (3) Plot no patterns
- Type option (1, 2, or 3):

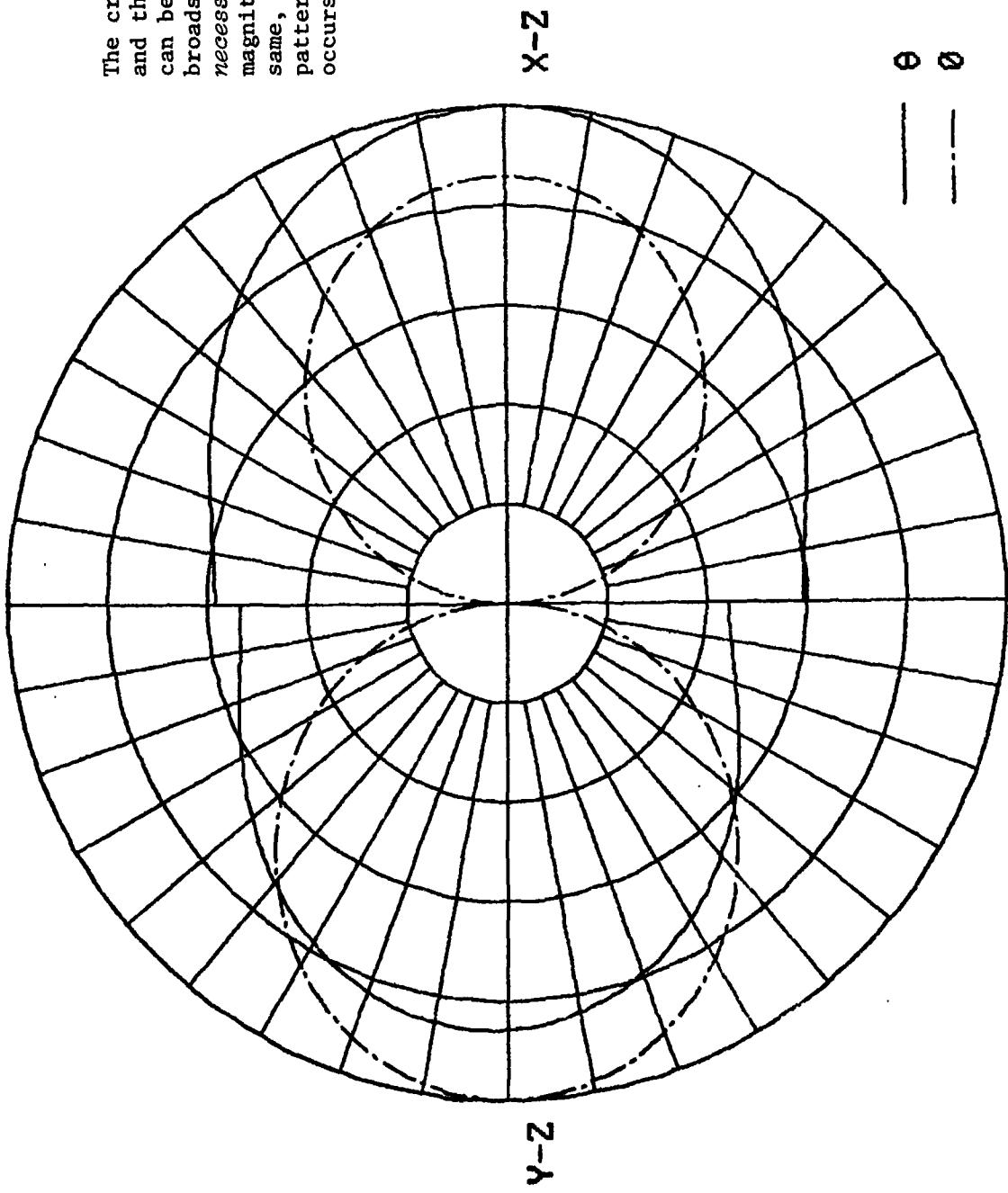
?1 INPUT CAPACITANCE IN PICOFARADS: C = ? 3.1

Notice that the band has been narrowed to 0.2% so that only a few patterns in the vicinity of the point at which CP is produced can be examined. To view all such patterns, the option "1" was selected.

CP OR LINEAR (TYPE C OR L) ? L

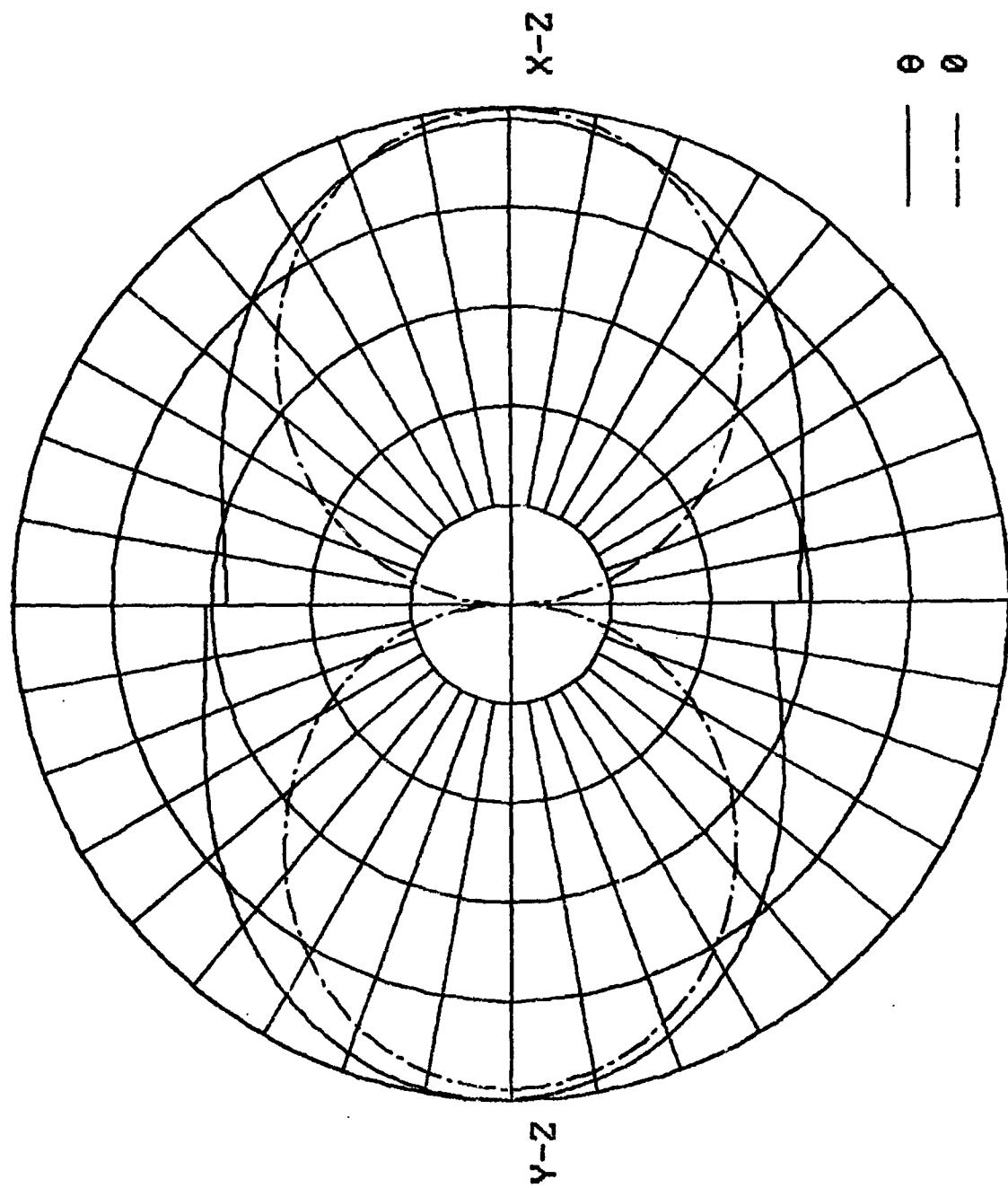
INDIVIDUAL NORMALIZATION (Y OR N) ? N

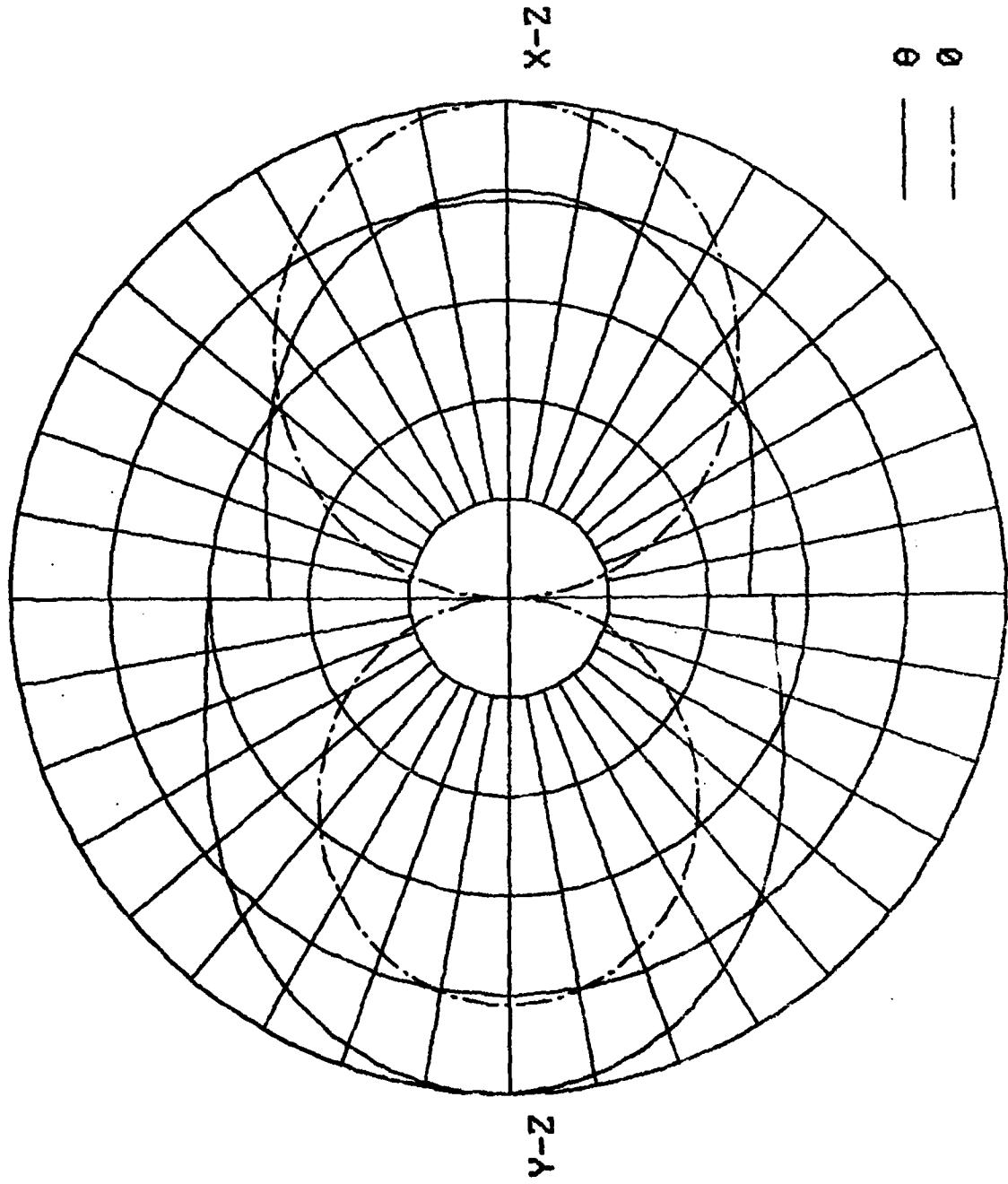
The cross-polarization in this and the subsequent patterns can be clearly seen. Since at broadside (the Z-direction) a necessary condition is that the magnitude of E_0 and E_θ be the same, a comparison of these patterns shows that the best CP occurs at $f = 1182 \text{ MHz}$.



F = 1182.0

80/03/11. 12.08.57.





80/03/11. 12.08.57.

F = 1183.0

RECTANGULAR, SMITH CHART, OR VS FREQUENCY PLOT (R, S, OR V)?

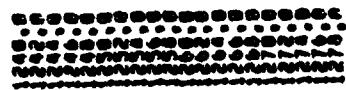
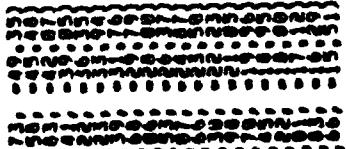
No impedance plot was needed since it was previously plotted and so a carriage return was input.

IN U POINT RECTANGULAR MICROSTRIP	
DIMENSIONS (A X B X T)	(7.62 X 7.68 X .15) CM
FED DATA:	
(1) WIDTH	1.25 CM
(2) PORT EDGE	1.45 CM
(3) PORT EDGE	3.98 CM
ELECTRICAL PARAMETERS:	
LOSS TANGENT	.0025
CONDUCTIVITY	21000000 KHMUS/CM
CENTER FREQUENCY	2.45 GHZ
FREQUENCY INCREMENT	.02 GHZ
QUALITY FACTOR	95.0
DATE / PAGE	08/03/11. 11.20.42.
FREQUENCY (MHZ)	2

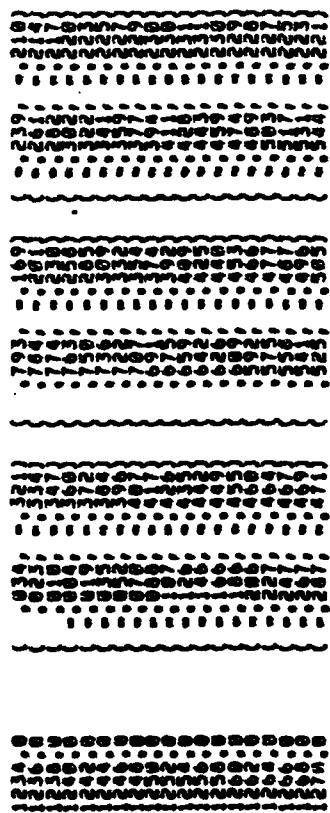
These computer printouts contain the geometrical and electrical parameters of the antenna, and the impedance or s-parameters in numerical form. In addition, the "directive gain" of the antenna (more precisely, the gain in the z direction) and the quality factor of the antenna are also printed. Output of this type is created for each antenna analyzed and is directed to file "RESULT." For brevity, however, only output for three of the cases considered in this chapter is included.

To associate a given printout with the graphical output, one simply matches the "date/time" group printed on both.

In this case, the "Z" column represents the driving point impedance of the one port.

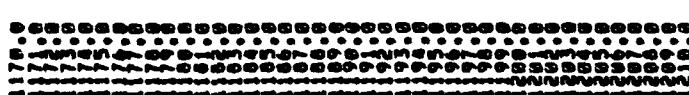
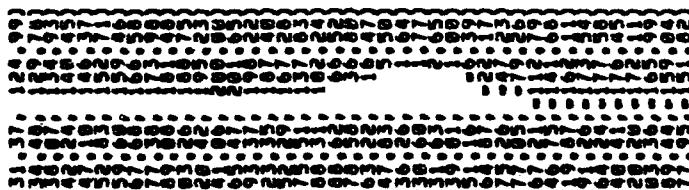


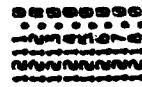
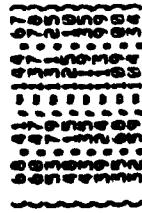
In this case, the s-parameters of the unloaded two-port are listed.



IN-PORT RECTANGULAR MICROSTRIP	
DIMENSIONS (A X B X T)	(7.62 X 7.69 X .15) CM
FEED DATA:	{ 2.5 C 1.88 } CM
WIDTH 1.50	HEIGHT 1.88
POINT 1.00	CAPACITIVE LOAD C = .14E+01
POW. THRU	
LOAD DESCRIPTION	
ELECTRICAL PARAMETERS	
REFLECTION LOSS	2.62195
LOSS TANDEL	2.98195
CONDUCTIVITY	21194.0 MHZ
CENTER FREQUENCY	99.6 MHZ
FREQUENCY RANGE	88.85 - 111.53.52.
QUALITY FACTOR	
SAR	
SAR/IAF	
FREQUENCY (MHZ)	Z

In this case, the input impedance of the capacitively loaded two-port is given under "Z."





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